

인공청각지능: 소리에서 의미로 Machine Listening: from sound to meaning

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Why Are We Here?



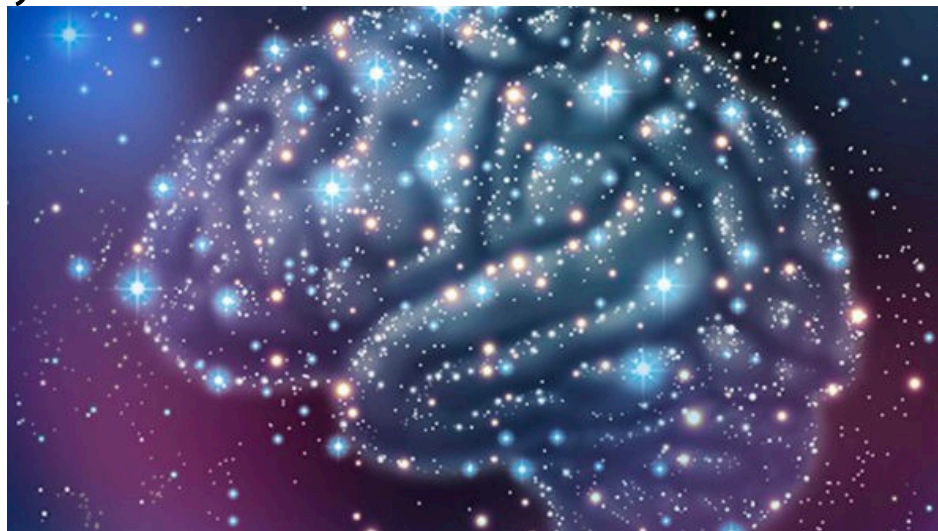
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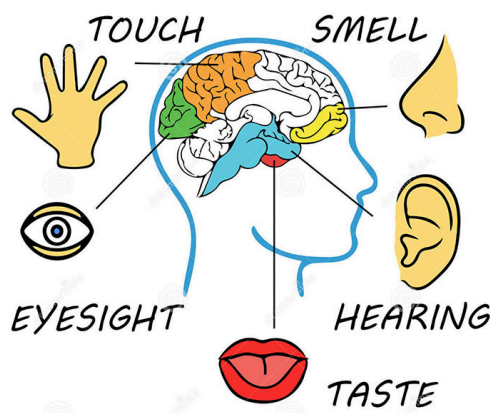
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Why Are We Here?

- We perceive and interact with the world through our senses.

- Sight
- Hearing
- Taste
- Smell
- Touch



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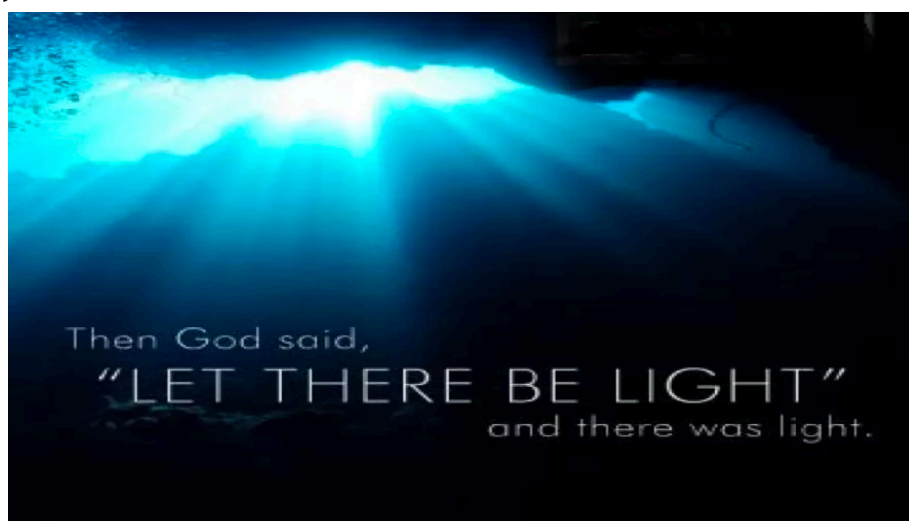
- Unfortunately, research in hearing (audition) is outnumbered by research in sight (vision) in every measure:
 - Number of researchers/scholars/students
 - Number of books/papers/articles
 - Number of academic communities
 - And more...

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Why Are We Here?



When you lose your vision, you lose
contact with things. When you lose
your hearing, you lose contact with
people.

— *Helen Keller* —

AZ QUOTES

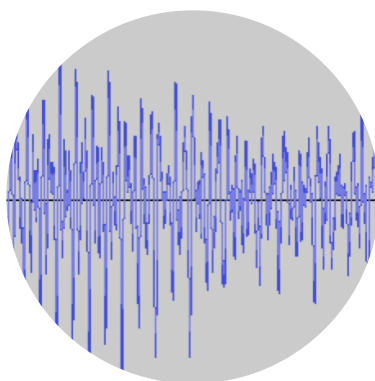
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Human Listening



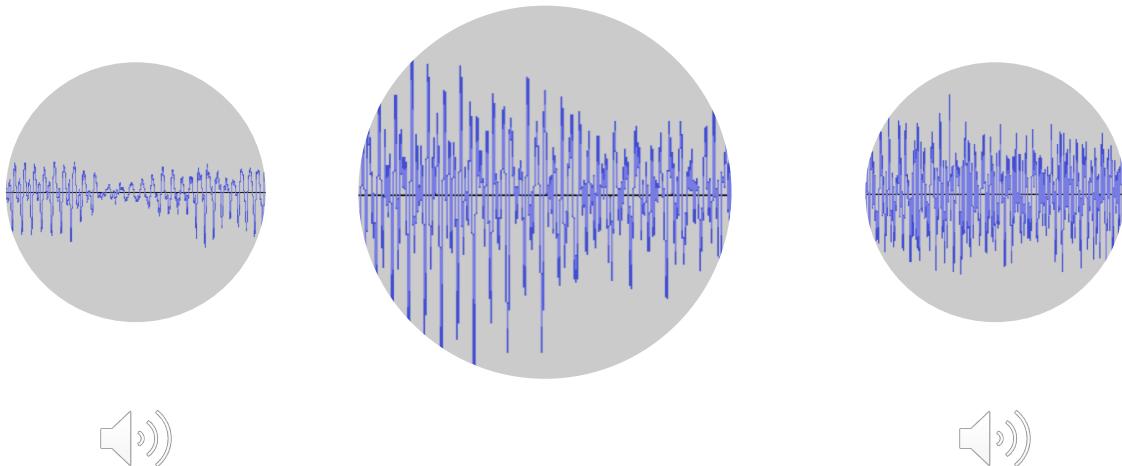
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Human Listening



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Human Listening

Arabic speech English speech Korean speech (Adult directed) Korean speech (Infant directed)



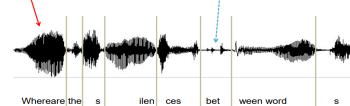
Where are the silences between words?



The segmentation problem

Where are the silences between words?

Where are the silences between words?

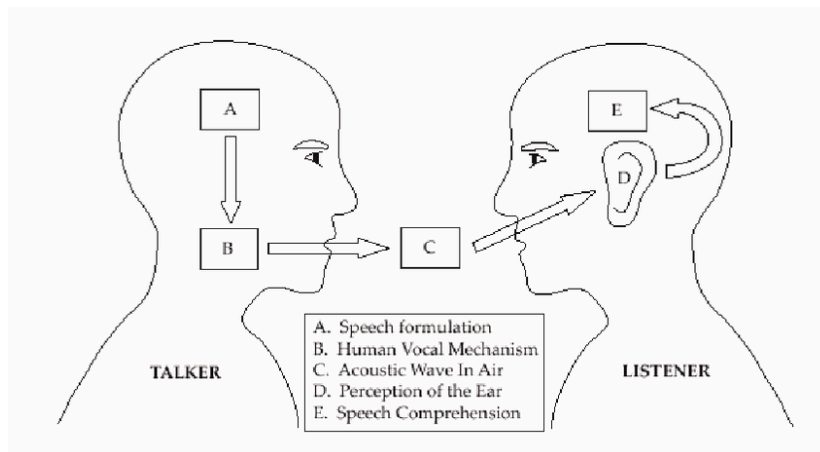


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Human Listening (& Speaking)

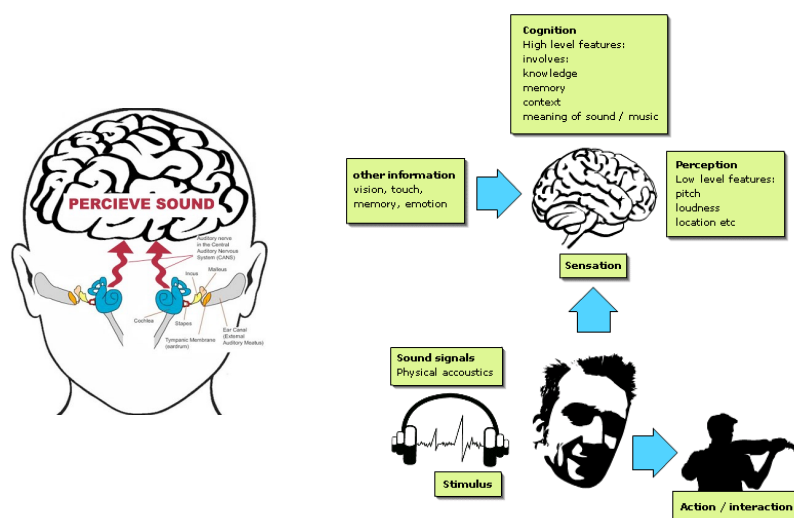


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Human Listening (& Speaking/Playing)

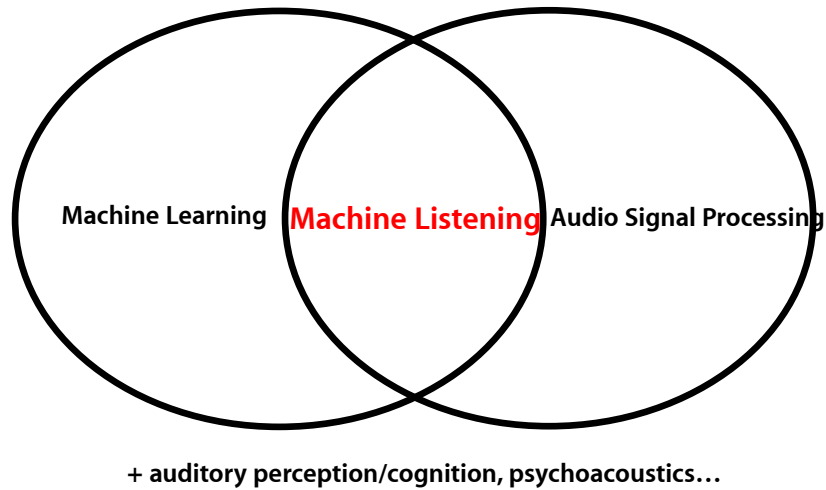


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Machine Listening (computer audition)



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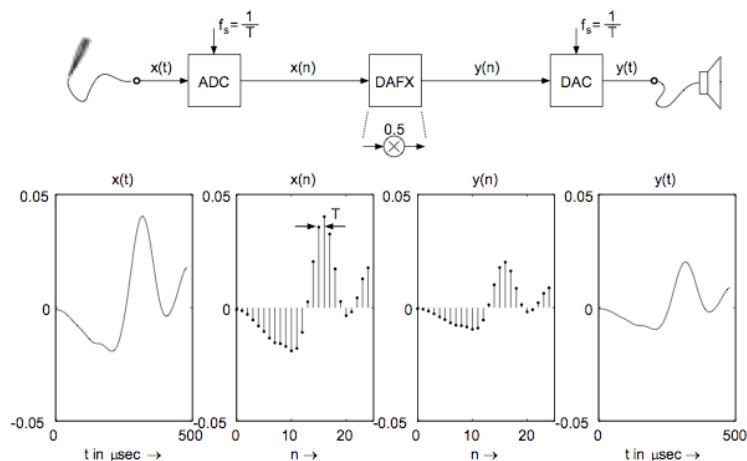
Fundamentals of Digital Audio Signal Processing

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Digital Signals



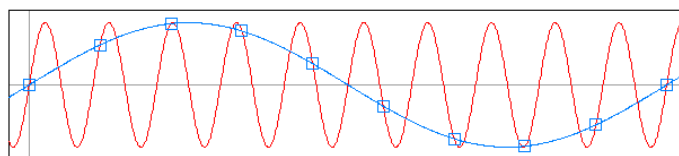
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Sampling

- According to the sampling theorem: $f_s > 2f_{max}$ (Nyquist limit)
- Otherwise there is another, lower-frequency, signal that share samples with the original signal (aliasing)



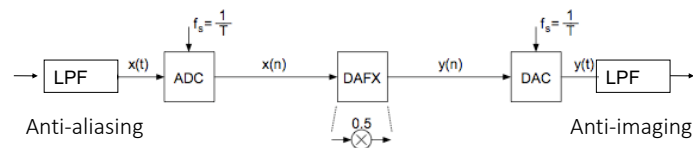
- Related to the wagon-wheel effect:
www.michaelbach.de/ot/mot_wagonWheel/index.html

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Anti-aliasing[imaging]



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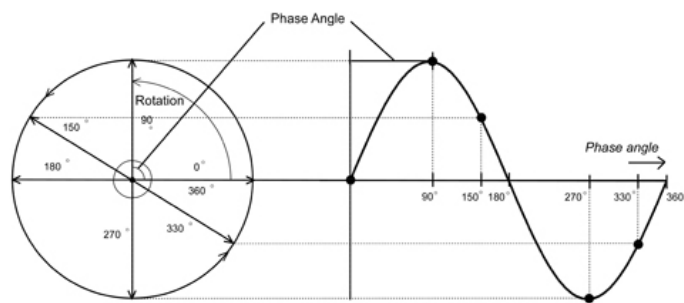
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Sinusoid (continuous-time)

$$x(t) = A \cos(\omega t + \phi) = A \cos(2\pi f t + \phi)$$

amplitude time (seconds) frequency (Hz)

angular frequency initial phase



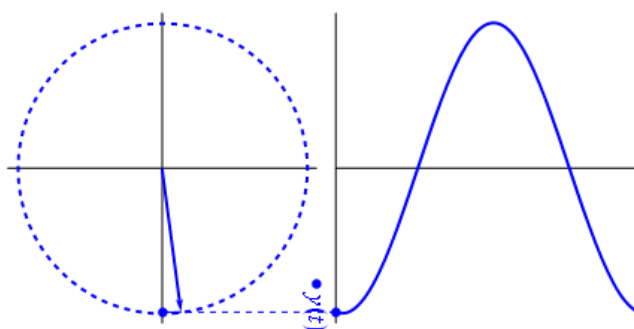
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Sinusoid (continuous-time) in action

$$x(t) = A \cos(\omega t + \phi) = A \cos(2\pi f t + \phi)$$



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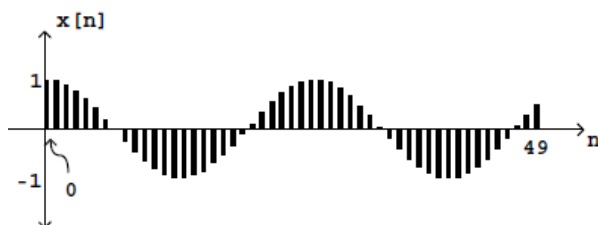
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Sinusoid (discrete-time)

- Since $t = \frac{n}{R}$, $x(n) = A \cos(\frac{2\pi f n}{R} + \phi)$,

where R = sampling rate



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Fourier Theorem

- Fourier theorem says: any *periodic* signal can be decomposed into a sum of sinusoids

$$x(t) = a_0 + \sum_{k=1}^{\infty} a_k \cos(2\pi kft + \phi_k)$$

- f is called *fundamental* and $2f, 3f, \dots$ are *harmonics* of fundamental
- Sequence of sinusoids with harmonic frequency is *harmonic series*

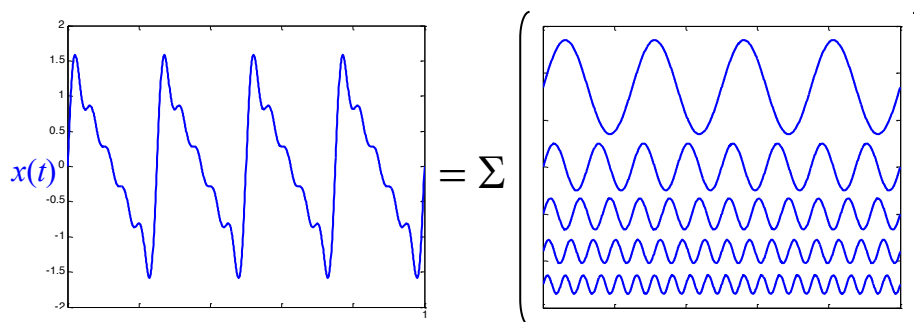


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Fourier Theorem

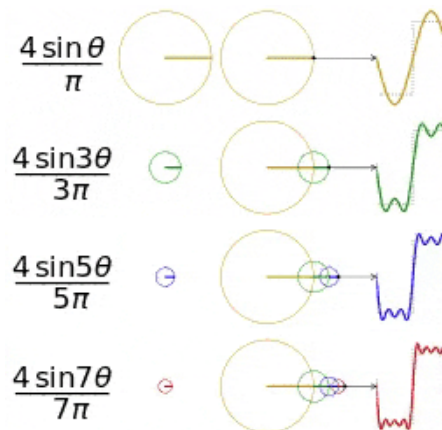


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Fourier Theorem



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Fourier Transform (FT)

- The Fourier transform of a continuous-time signal $x(t)$ may be defined as

$$X(\omega) = FT[x(t)] = \int_{-\infty}^{\infty} x(t) e^{-j\omega t} dt, \quad \omega \in (-\infty, \infty)$$

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Discrete Fourier Transform (DFT)

- The Discrete Fourier Transform of a signal $x(n)$ may be defined as

$$X(k) = DFT[x(n)] = \sum_{n=0}^{N-1} x(n)e^{-j2\pi nk/N}, \quad k = 0, 1, \dots, N-1$$

- The resulting N samples $X(k)$ are complex-valued:

$$X(k) = X_R(k) + jX_I(k)$$

$$|X(k)| = \sqrt{X_R^2(k) + X_I^2(k)} \quad : \text{magnitude}$$

$$\varphi(k) = \arctan \frac{X_I(k)}{X_R(k)} \quad : \text{phase}$$

$$k = 0, 1, \dots, N-1$$

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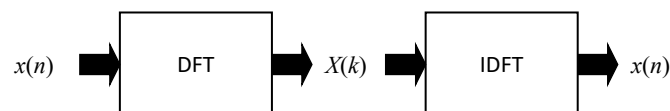
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Inverse DFT (IDFT)

- The DFT allows perfect reconstruction of a signal $x(n)$ from its DFT $X(k)$ via inverse DFT defined as:

$$x(n) = IDFT[X(k)] = \frac{1}{N} \sum_{k=0}^{N-1} X(k)e^{j2\pi nk/N}, \quad n = 0, 1, \dots, N-1$$



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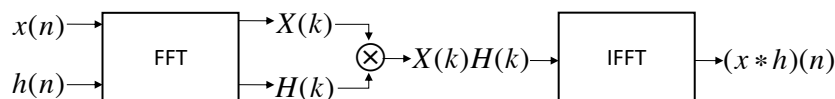
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Fast Fourier Transform (FFT)

- An algorithm to efficiently compute the DFT is known as the Fast Fourier Transform (FFT) and its inverse as the IFFT
- Computational complexity
 - N -point DFT: $O(N^2)$
 - N -point FFT: $O(N \log N)$
- The FFT is so fast that even time-domain operations, like convolution, can be performed faster using FFT and IFFT instead:

$$(x * h)(n) = \sum_{m=0}^{N-1} x(m)h(n-m)$$

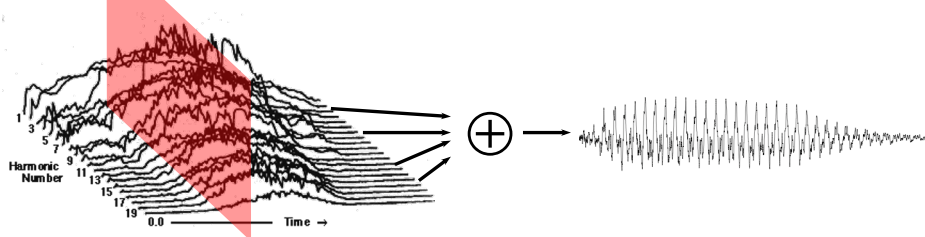


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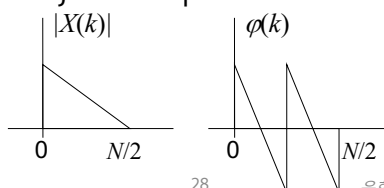
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Revisiting the Fourier Theorem



- Any periodic signal can be described by a sum of a series of sinusoids with time-varying amplitudes and phases
- Thus a complex spectrum is just a snapshot of those sinusoids' parameters

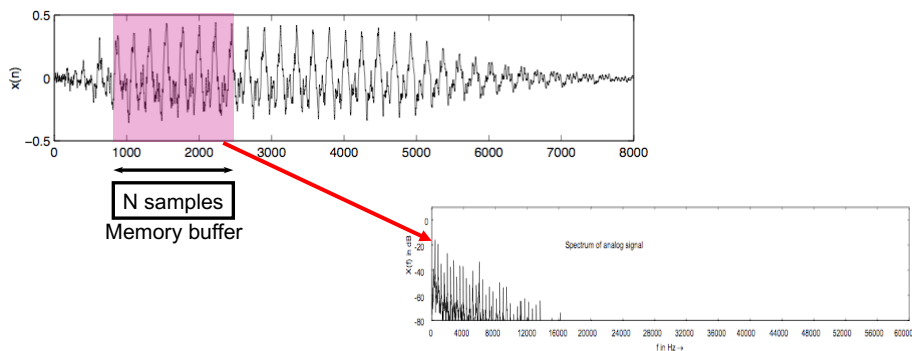


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Short-time Fourier Transform (STFT)



- For block processing, a short segment is sent to a buffer and processed as a block
- The DFT done in this way is called the short-time Fourier Transform or STFT

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Frequency Resolution

- Determined by how many sinusoids are used to describe a spectrum
- In N -point DFT or FFT, the frequency resolution is given by

$$\Delta f = f_s / N$$

- Intuitively, we can have finer frequency resolution if we increase N
- However, this results in poorer temporal resolution => tradeoff between frequency[spectral] and time[temporal] resolution

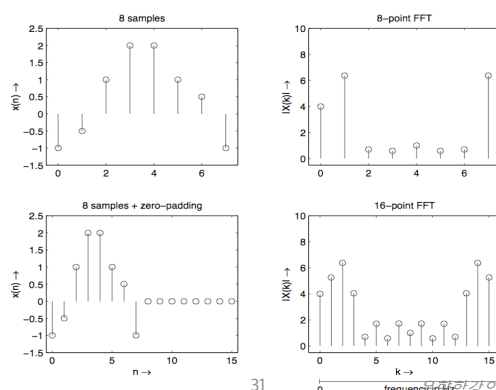
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Zero-Padding

- A possible solution to increase frequency resolution without increasing N - i.e., without losing time resolution
- Add zero-valued samples - thus doesn't change spectrum itself - to yield better spectral resolution

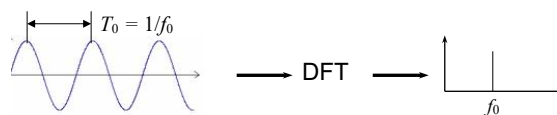


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DFT of a Sinusoid

- In theory the DFT of a pure sinusoid results in a single sharp line at the frequency of the sinusoid



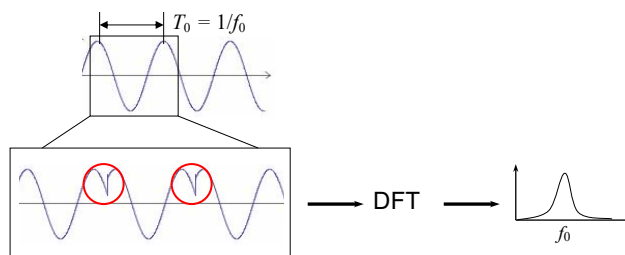
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Spectral Leaking

- In practice, unless we perform f_0 -synchronous analysis, there are discontinuities (sharp changes) at the segment boundaries that introduce some noise. Thus the spectral line around f_0 is smeared.
- This is known as *spectral leaking*



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Windowing

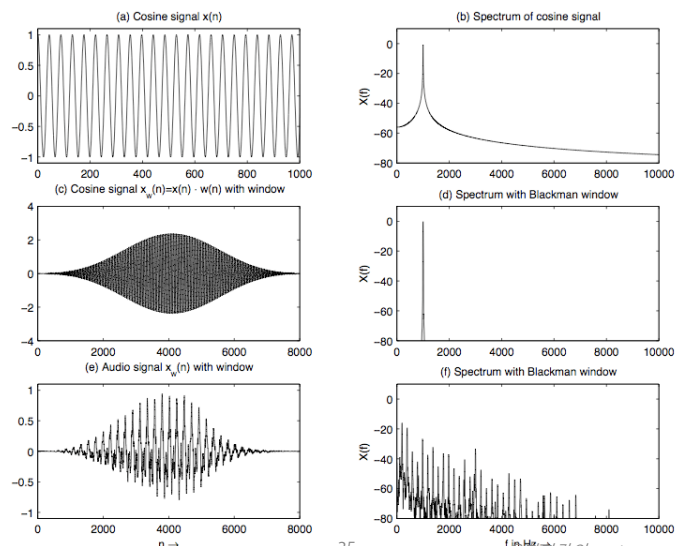
- In order to reduce spectral leaking, we need to avoid abrupt changes between the segment boundaries
- This is done by multiplying a *window* whose amplitude gradually reaches zero at both ends, thus guaranteeing the continuity of a segmented signal when repeated
- Possible windows are: rectangular, hamming, hann, Blackman, Gaussian, and so on.

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Windowing Example



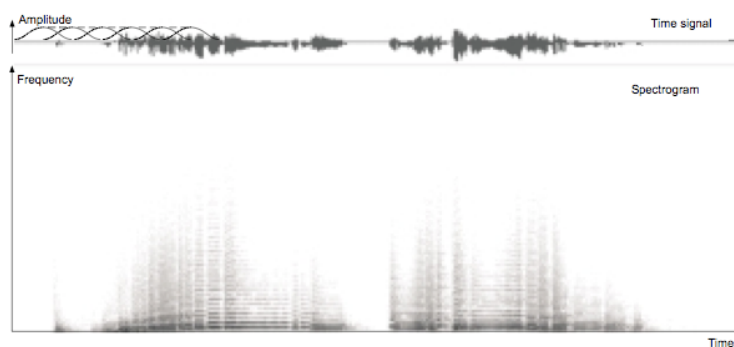
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2-D Time-Frequency Representation

- Using the STFT, independent DFTs are calculated on windowed segments
- The segments usually overlap to compensate for the loss of temporal resolution
- Produces a 2-D *spectrogram*



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Acoustic Features

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Audio Feature Extraction

- Raw audio samples are:
 - Noisy
 - Redundant
 - Computationally inefficient
 - Not a good input to audio applications
- Need to convert them to more robust, compact yet meaningful representations called *audio features* or *acoustic features*

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Acoustic (audio) Features

- Spectral features
 - Spectral low-level features: spectral centroid, spectral flatness measure, spectral flux, etc.
 - Spectral envelope: LPCs, MFCCs, etc.
- Temporal features
 - ZCR (zero crossing rate), tempo histogram, novelty function, etc.
- Tonal features
 - PCP (pitch class profile) or chroma, chromagram, tonal centroid, etc.

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Spectral Low-level Features

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Spectral Centroid (SC)

$$SC = \frac{\sum_{k=0}^{N/2} f_k |X(k)|^2}{\sum_{k=0}^{N/2} |X(k)|^2},$$

where f_k is the center frequency of k th bin and $|X(k)|$ is the DFT

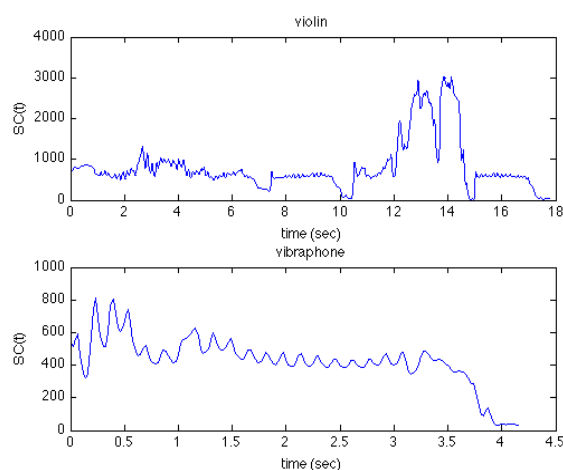
- Characterizes the center of gravity of the (power) spectrum
- Usually associated with the “sharpness / dullness (or brightness / darkness)” of a sound

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Spectral Centroid: Example



(a) violin



(b) vibraphone



Temporal evolution of the spectral centroid for two instrumental sounds

(a) violin and (b) vibraphone

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Spectral Spread (SS)

$$SS = \sqrt{\frac{\sum_{k=0}^{N/2} (f_k - SC)^2 |X(k)|^2}{\sum_{k=0}^{N/2} |X(k)|^2}}$$

- Measure of the average spread of the spectrum in relation to its centroid
- Noisy, broadband signals have high SS while tonal sounds show lower SS

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Spectral Flatness Measure (SFM)

$$SFM_b = \frac{\sqrt[N_b]{\prod_{k_b} |X(k_b)|^2}}{\frac{1}{N_b} \sum_{k_b} |X(k_b)|^2}, \quad k_b = k_l, k_l + 1, \dots, k_u$$

where N_b is the number of spectral bins in a subband or
 $N_b = k_u - k_l + 1$

- Reflects how "flat" a signal's power spectrum is
- Calculated as the ratio of the geometric mean and the harmonic mean
- Usually computed *per* spectral band (critical bands or bark bands, etc.), thus SFM_b
- Flatness for the whole spectrum is the average of the subband measures

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Harmonic Spectral Centroid (HSC)

$$HSC = \frac{\sum_{h=1}^{N_h} f_h A_h}{\sum_{h=1}^{N_h} A_h}$$

where f_h and A_h are the frequency and the amplitude of the h_{th} harmonic, respectively

- Measure of the amplitude-weighted mean of the harmonic (spectral) peaks of the spectrum
- Compared to SC, HSC focuses only on harmonic (spectral) peaks, which are more musically meaningful in general
- Harmonic spectral spread (HSS) is similarly defined

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Spectral Envelope

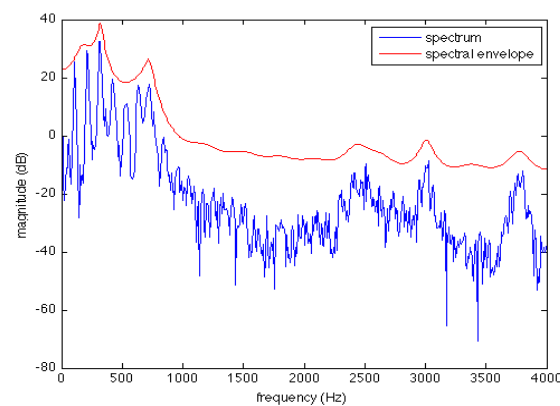
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What is the Spectral Envelope?

- Spectral envelope is a smoothed version of the spectrum that preserves its overall shape while neglecting its fine spectral structure

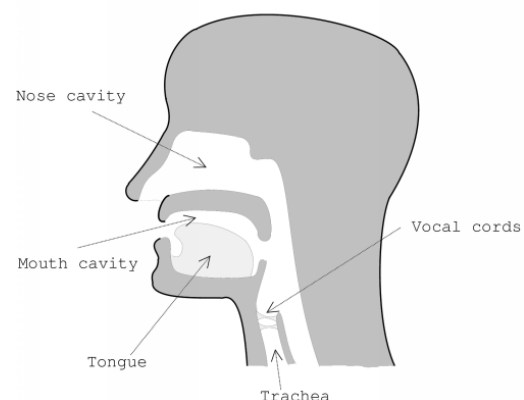


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Human Speech System

- Vocal cords act as an oscillator, which generates a spectrally rich *source* signal
- Everything else is *filter*: vocal tract, mouth/nose cavity, tongue
- Thus called “source-filter” model
- These filters define the shape of the spectral envelope



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Spectral Envelope Estimation

- A few popular techniques to estimate the spectral envelope
 - Channel vocoder: estimates the amplitude of the signal within several frequency bands
 - Linear prediction: estimates the parameters (or filter coefficients) of a filter that approximates the spectrum
 - Cepstrum analysis: inverse-FFT the log-spectrum and low-pass filters it to obtain the envelope

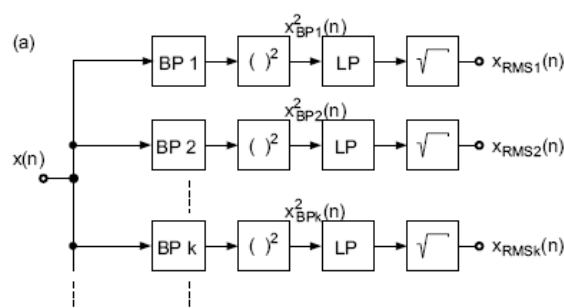
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Channel Vocoder (1)

- Filters a signal with a bank of bandpass filters
- Calculates RMS of each bandpassed signal
- The more filters used, the finer spectral envelope estimated



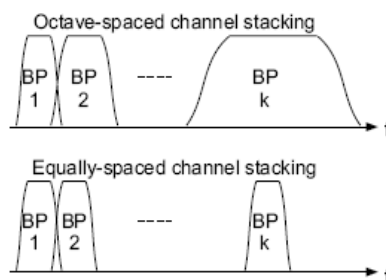
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Channel Vocoder (2)

- In the frequency domain, multiply the spectrum with the filters' frequency response and square-root the sum of each filter's output



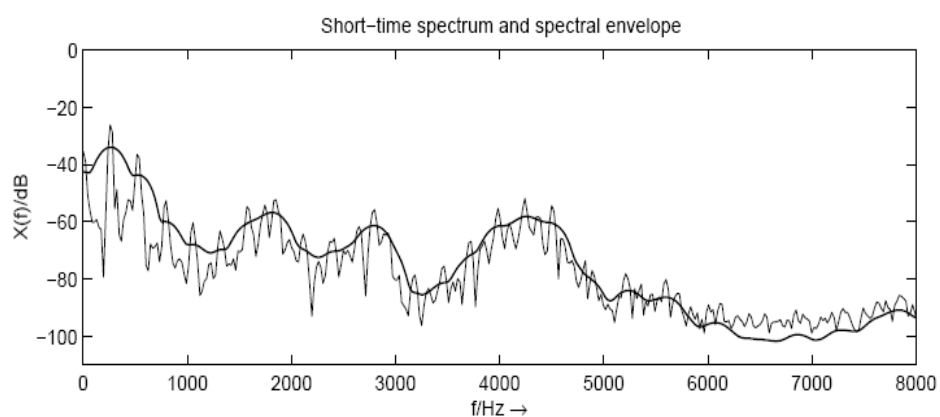
- The filterbank can be either linearly or logarithmically spaced (e.g., constant-Q or Mel-scale filter bank)

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Channel Vocoder Example



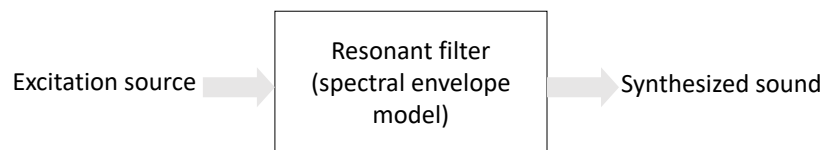
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Linear Predictive Coding (1)

- Linear predictive coding is a source-filter model that approximates the way a sound is generated as an excitation (a pulse train or noise) passing through an all-pole resonant filter



- Widely used in speech and music applications
- Reduces large amount of data (e.g., N samples) to a few filter coefficients while preserving the overall shape

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Linear Predictive Coding (2)

- The n th sample $x(n)$ is extrapolated, i.e., predicted by a linear combination of p past samples:

$$x(n) \approx \hat{x}(n) = \sum_{k=1}^p a_k x(n-k)$$

- The residual error is given by

$$e(n) = x(n) - \hat{x}(n) = x(n) - \sum_{k=1}^p a_k x(n-k)$$

and we want to minimize this error

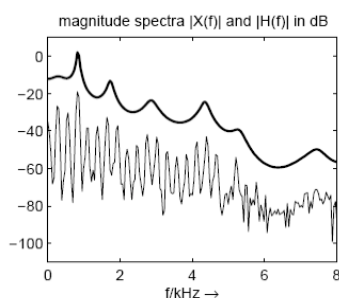
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Linear Predictive Coding (3)

- The parameters a_k 's are called linear predictive coefficients (LPCs)
- The filter represented by these coefficients is a resonant filter and its frequency response represents the spectral envelope
- The higher the filter order p , the closer the approximation is to the signal's spectrum

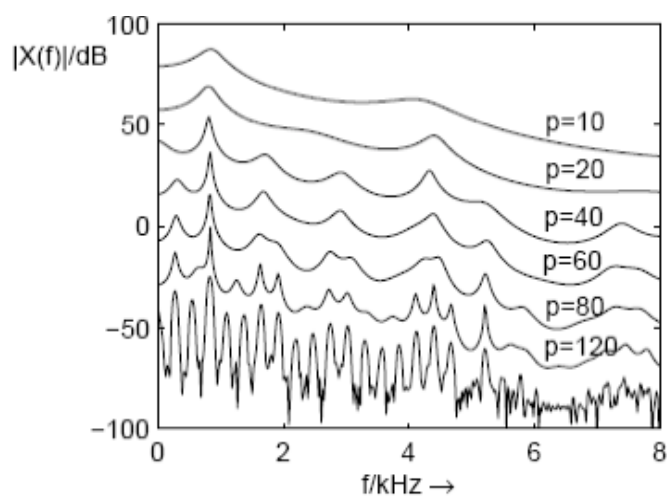


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LPC Order and Approximation



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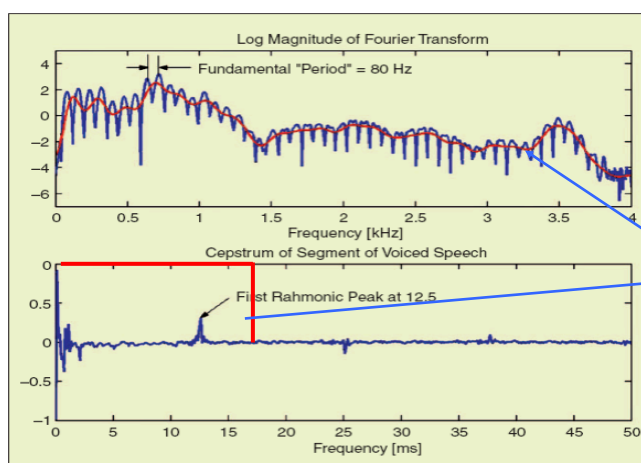
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Cepstrum Analysis

- Cepstrum is the result of taking the FFT of the log-spectrum as if it were a *signal*
- Measures the rate of change in different spectral bands
- The name cepstrum was coined by Bogert *et al.* (1963) by reversing the first four letters of the spectrum (similarly for quefrency analysis and liftering, etc.)
- For a real signal $x(n)$, the real cepstrum is calculated as follows:

$$c_R(n) = \text{IFFT}(\log(|X(k)|))$$

Cepstrum & Spectral Envelope



Spectral Envelope Estimation

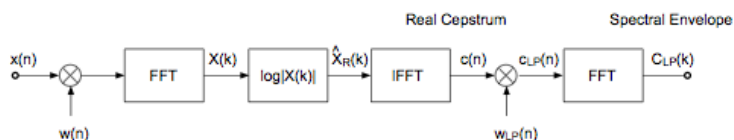
- Using a low-pass window of the form:

$$\omega_{LP}(n) = \begin{cases} 1 & n = 0, N_1 \\ 2 & 1 \leq n \leq N_1 \\ 0 & N_1 < n \leq N-1 \end{cases}$$

we can low-pass the cepstrum and obtain the spectral envelope by:

$$c_{LP}(n) = c_R(n) \cdot \omega_{LP}(n)$$

$$C_{LP}(k) = FFT[c_{LP}(n)]$$

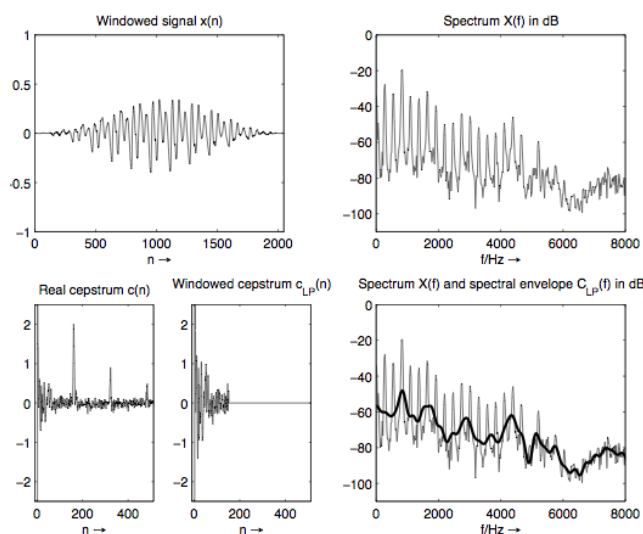


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Cepstrum Example



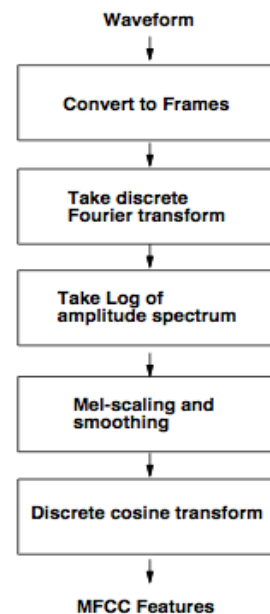
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MFCC

- Mel-frequency Cepstral Coefficients (MFCCs) are a variation of cepstrum, motivated by human perception (Logan, 2000)
- Most extensively used in speech and music applications (e.g., speech recognition, genre classification, instrument recognition, etc.), due to its ability to compactly represent the spectral characteristics (just ~13 coefficients)



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Mel Scale

- The Mel scale is a non-linear perceptual scale of pitches judged to be equidistant
- Approximately linear below 1 kHz and logarithmic above
- 1 kHz corresponds to 1000 Mel (reference point)
- With the Mel scale, a 1000-Mel tone should sound as twice as high as a 500-Mel tone (this is not true with linear frequency Hz)

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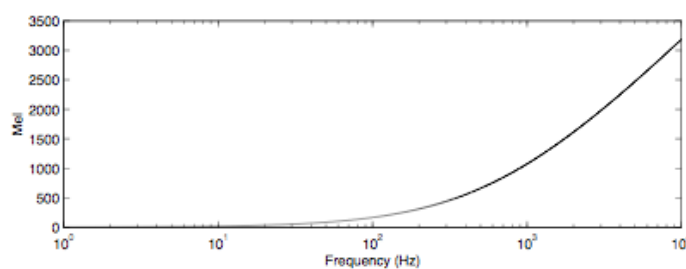
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Mel vs. Linear Frequency

- The relation between Mel and Hz is given by

$$m = 1127.01048 \log(1 + f / 700)$$

$$f = 700(e^{m/1127.01048} - 1)$$



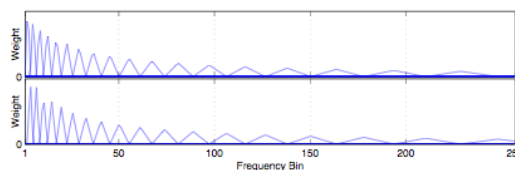
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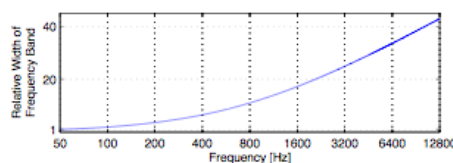
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Mel-frequency Spectrum

- To convert a linear spectrum to Mel we can use a filterbank of overlapping triangular windows:



- Such that the width d of each window increases according to the Mel scale, and the height of each triangle is $2/d$



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Decorrelation of Mel-scale Spectrum

- The resulting Mel-scale spectral vectors are highly correlated with each other; i.e. highly redundant
- Thus a more efficient representation of the log-spectrum can be obtained by applying a transform that decorrelates those vectors (Rabiner and Juang, 93)
- This decorrelation is commonly approximated by means of the Discrete Cosine Transform (DCT)
- The DCT is similar to a DFT but only for real numbers. It has the property that most of its energy is concentrated on a few initial coefficients (thus effectively compressing the spectral info)

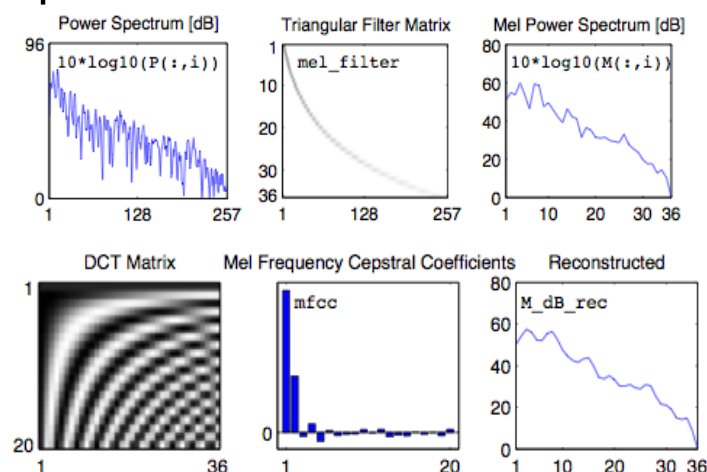
$$X_{DCT}(k) = \sqrt{\frac{2}{N}} \sum_{n=0}^{N-1} x(n) \cos\left[\frac{\pi}{N}\left(n + \frac{1}{2}\right)k\right]$$

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Fast Computation of MFCCs



MFCCs roughly model certain characteristics of human auditory perception: the non-linear perception of loudness and frequency and spectral masking (Pampalk, 2006)

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Tonal Features

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Tonality

- Very important attribute in (Western) tonal music
- Explain the relationship among the tones
- Several musical attributes are closely related with tonality
 - Scale
 - Key
 - Pitch
 - Interval
 - Chord

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Tonal Features vs. Spectral Features

- Spectral features
 - Good for describing certain spectral characteristics (e.g., sharpness, noisiness, etc.)
 - Good for representing sonic texture or timbre by capturing overall frequency magnitude response (e.g., LPCs, cepstral coefficients, MFCCs)
 - Not good for tonal analysis: pitch- or tone-relevant information gets lost
- Tonal features retain tonal structure in musical audio
 - Tonal relations
 - Interval relations

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Constant-Q Transform

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Constant-Q Transform

- In DFT, the center frequency f_k of the frequency bin is given by

$$f_k = \frac{f_s}{N} k, \quad k = 0, 1, \dots, N-1$$

where f_s is the sampling rate and N is the DFT size

- Therefore, all the frequency bins are linearly spaced
- However, musical scale as well as human hearing mechanism are *logarithmic*
- Brown proposed the constant-Q transform whose frequency resolution conforms to equal-tempered scale (1990)
- Well suited for pitch-related analysis

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Constant-Q Transform (cont'd)

- In constant-Q transform the k th spectral frequency is defined as

$$f_k = (2^{1/B})^k f_{\min}, \quad k = 0, 1, \dots, N-1,$$

where B is the number of bins in an octave and $f_{\min}$ is the minimum frequency set by user

- It is called "constant-Q" because Q or "quality factor" is constant along the frequency axis, which is defined as

$$Q = \frac{f_k}{f_w}, \quad k = 0, 1, \dots, N-1,$$

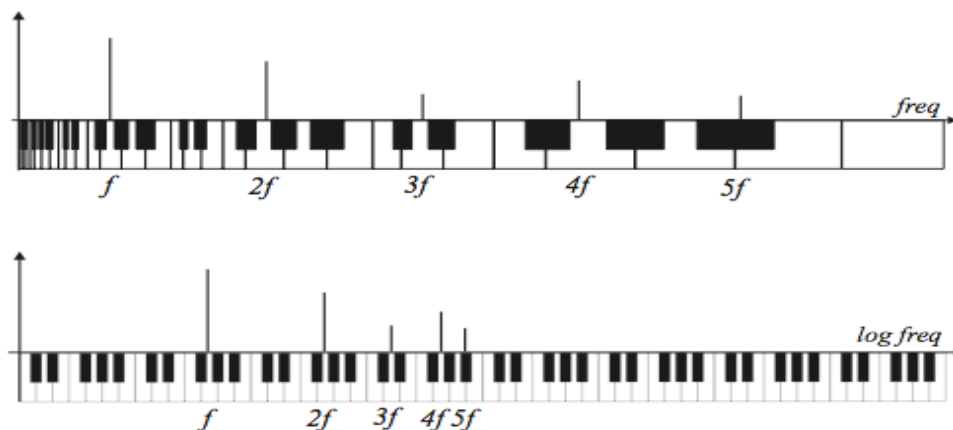
where f_w is the filter width

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Linear vs. Logarithmic Frequency



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Computation of CQ Transform

- CQ transform can be obtained from the DFT using logarithmically-spaced filterbank

$$X_{cq}(k) = \frac{1}{N(k)} \sum_{n=0}^{N(k)-1} x(n)w(n,k)e^{-j2\pi Qn/N(k)}$$

$$N(k) = f_s Q / f_k$$

- That uses a variable window length to obtain more resolution at lower frequencies and less at higher (logarithmic distribution of bins in frequency)

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Computation of CQT (cont'd)

- CQ transform can be efficiently computed using a *CQ kernel* which is a 2-d matrix that maps the DFT to the CQT

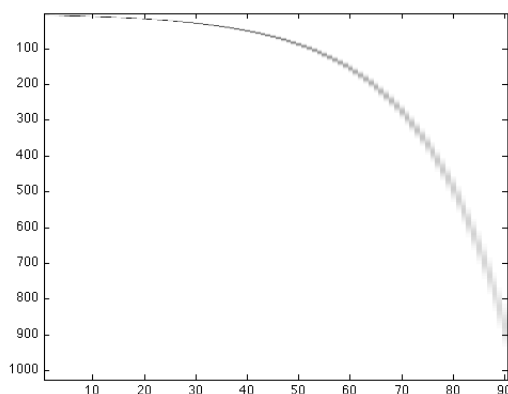
$$\boxed{\text{N-point DFT vector}} \times \boxed{\begin{array}{c} \text{NxM} \\ \text{CQ kernel} \\ \text{matrix} \end{array}} = \boxed{\text{M-point CQT vector}}$$

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CQ Kernel Matrix



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Chroma

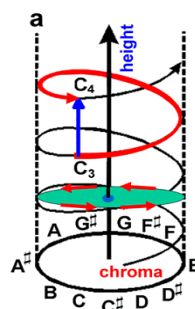
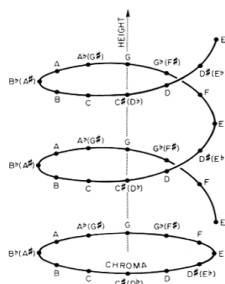
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Pitch Helix

- The pitch helix is a pitch space where linear pitch is wrapped around a cylinder, thus modeling the special relationship that exists between octave intervals



- Two dimensions
 - Height: naturally organizes *absolute* pitches from low to high
 - Chroma: represents the inherent circularity of pitch (*relative* relationship between pitch classes)

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Chroma (aka Pitch Class Profile)

- Good for describing relative pitch relationship, disregarding absolute pitch height
- Very useful for harmony analysis, key and chord, in particular
- A key and/or a chord can be described as a function of its pitch classes
- Almost universal feature for key/chord estimation applications
- Chroma as audio feature first introduced by Fujishima (1999)

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Computation of Chroma

- Easily computed from the constant-Q transform by collapsing it to an octave, or

$$Chroma(b) = \sum_{m=0}^{M-1} |X_{CQ}(b + mB)|,$$

where $X_{CQ}(k)$ is the CQ transform,

M is the total number of octaves of interest,

B is the number of chroma bins in an octave,

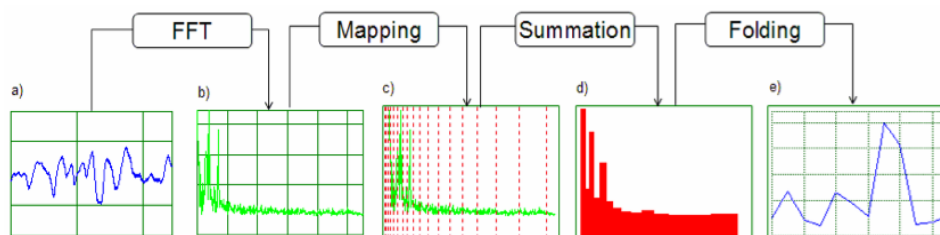
and $b = 1, 2, \dots, B$ is the chroma bin index

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Computation of Chroma (cont'd)

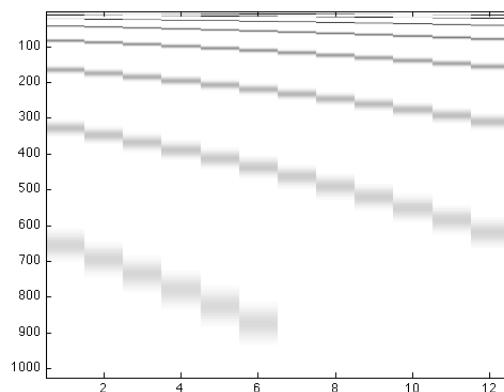


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Chroma Kernel Matrix



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Machine Learning

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Machine Learning definition

- Arthur Samuel (1959). Machine Learning: Field of study that gives computers the ability to learn without being explicitly programmed.
- Tom Mitchell (1998) Well-posed Learning Problem: A computer program is said to *learn* from experience E with respect to some task T and some performance measure P , if its performance on T , as measured by P , improves with experience E .

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Machine learning algorithms:

- Supervised learning
- Unsupervised learning

Others: Reinforcement learning, recommender systems.

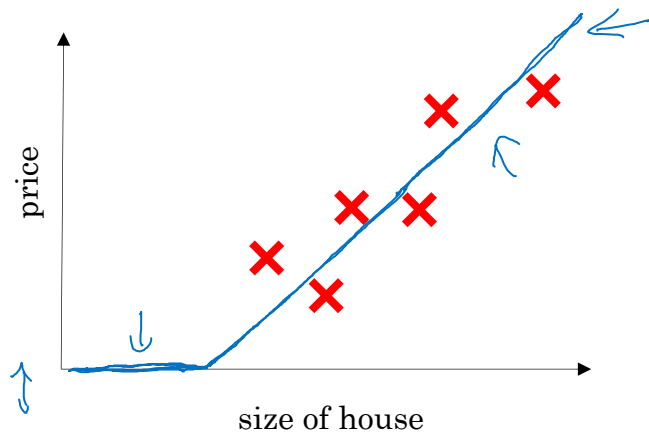


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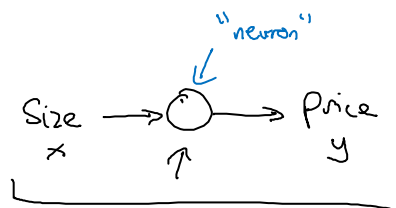
Introduction to Deep Learning

What is a Neural Network?

Housing Price Prediction



ReLU
Rectified
Linear
Unit

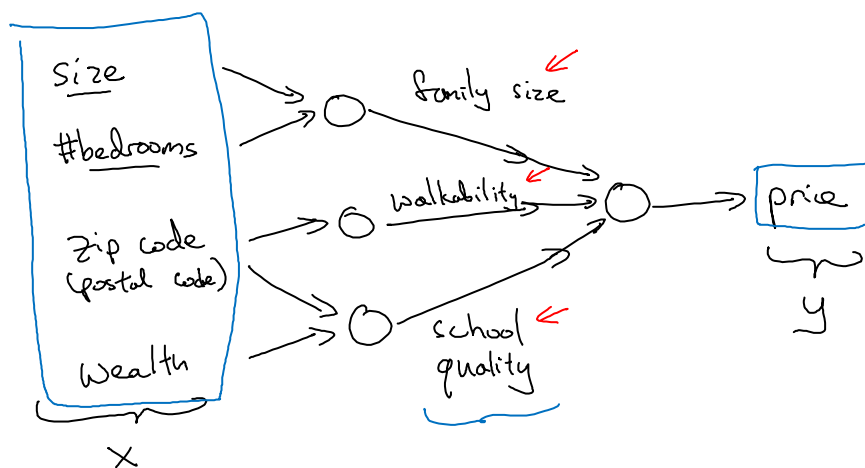


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Housing Price Prediction



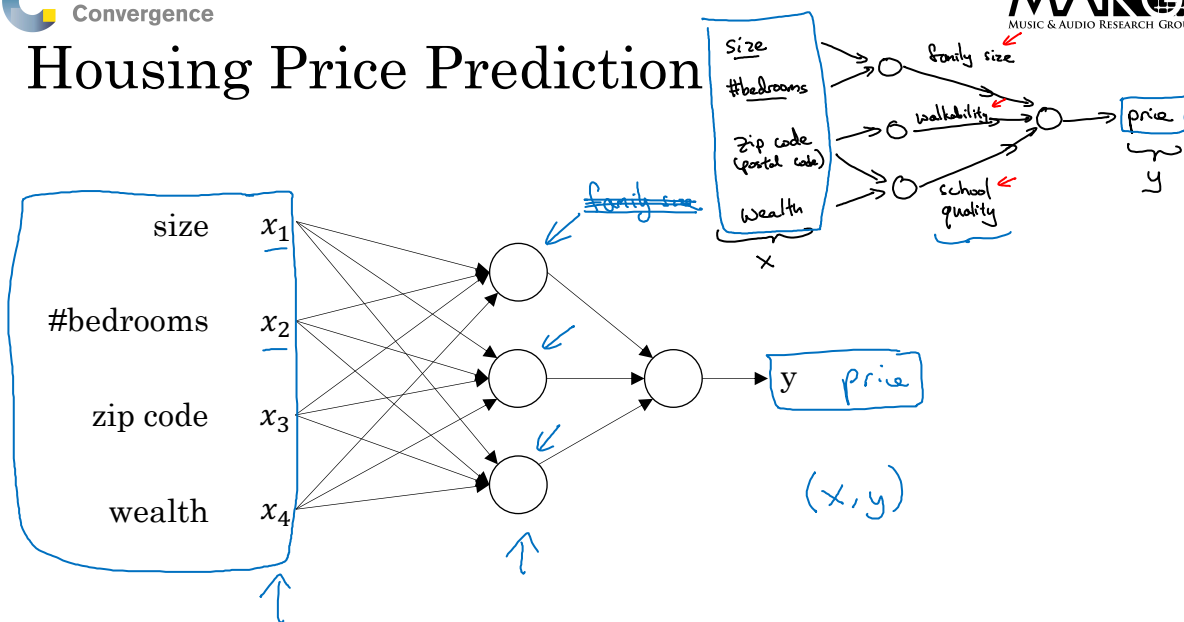
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Housing Price Prediction



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Introduction to Deep Learning

Supervised Learning with Neural Networks

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Supervised Learning

Input(x) 	Output (y) 	Application
Home features	Price	Real Estate
Ad, user info 	Click on ad? (0/1)	Online Advertising
Image	Object (1,...,1000)	Photo tagging
<u>Audio</u>	Text transcript	Speech recognition
<u>English</u>	Chinese	Machine translation
<u>Image, Radar info</u> 	<u>Position of other cars</u> 	Autonomous driving

Standard NN
CNN
RNN
Custom/Hybrid

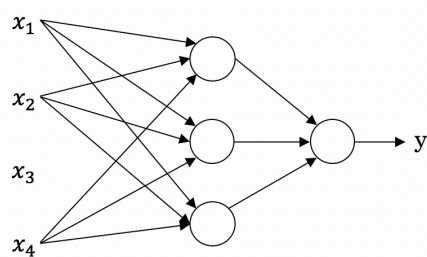
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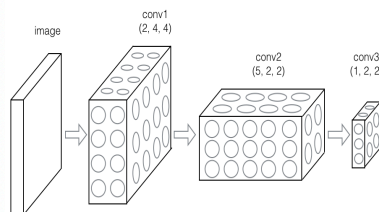
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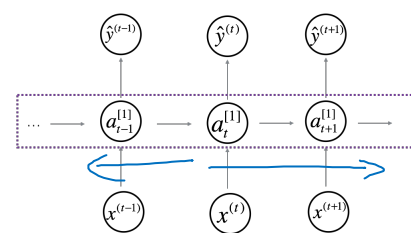
Neural Network examples



Standard NN



Convolutional NN



Recurrent NN

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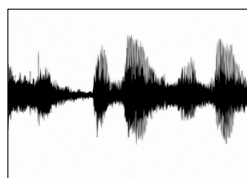
Supervised Learning

Structured Data

Size	#bedrooms	...	Price (1000\$s)
2104	3		400
1600	3		330
2400	3		369
⋮	⋮		⋮
3000	4		540

User Age	Ad Id	...	Click
41	93242		1
80	93287		0
18	87312		1
⋮	⋮		⋮
27	71244		1

Unstructured Data



Audio



Image

Four scores and seven
years ago...

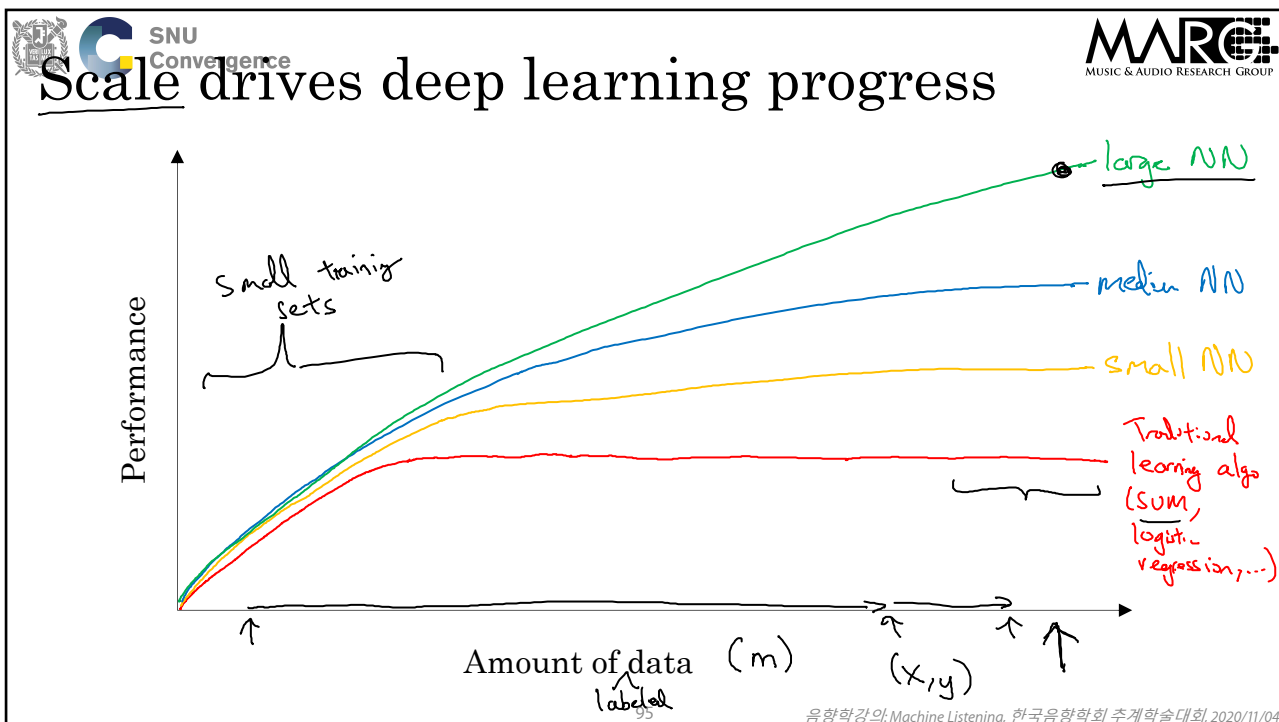
Text



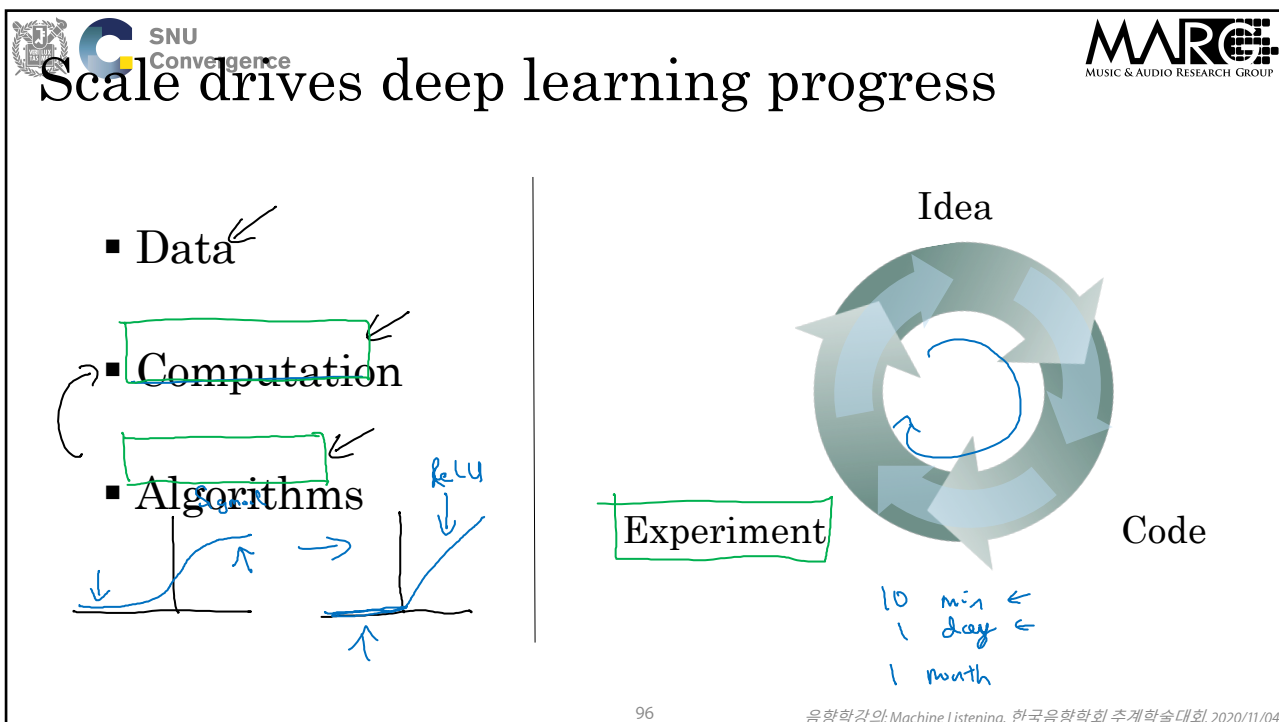
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Introduction to Neural Networks

Why is Deep Learning taking off?



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Basics of Neural Network Programming

Binary Classification

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SNU

Binary Classification

MUSIC & AUDIO RESEARCH GROUP

64

64

→

1 (cat) vs 0 (non cat)

y

64

64

$X =$

$$\begin{bmatrix} 255 \\ 231 \\ \vdots \\ 255 \\ 134 \\ \vdots \end{bmatrix}$$

$64 \times 64 \times 3 = 12288$

$n = n_x = 12288$

$X \rightarrow y$

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Notation

$$(x, y) \quad x \in \mathbb{R}^{n_x}, y \in \{0, 1\}$$

$$m \text{ training examples} : \{(\underline{x}^{(1)}, \underline{y}^{(1)}), (\underline{x}^{(2)}, \underline{y}^{(2)}), \dots, (\underline{x}^{(m)}, \underline{y}^{(m)})\}$$

$$M = M_{\text{train}}$$

$$M_{\text{test}} = \# \text{test examples.}$$

$$X = \begin{bmatrix} | & | & & | \\ x^{(1)} & x^{(2)} & \dots & x^{(m)} \\ | & | & & | \end{bmatrix}$$

$X \in \mathbb{R}^{n_x \times m}$ $X.\text{shape} = (n_x, m)$

$$Y = [y^{(1)} \ y^{(2)} \ \dots \ y^{(m)}]$$

$$Y \in \mathbb{R}^{1 \times m}$$

$$Y.\text{shape} = (1, m)$$



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Basics of Neural Network Programming

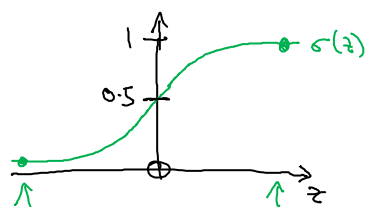
Logistic Regression

Logistic Regression

Given x , want $\hat{y} = P(y=1|x)$
 $x \in \mathbb{R}^{n_x}$
 $0 \leq \hat{y} \leq 1$

Parameters: $\underline{w} \in \mathbb{R}^{n_x}$, $\underline{b} \in \mathbb{R}$.

Output $\hat{y} = \sigma(\underbrace{w^T x + b}_z)$



$$x_0 = 1, \quad x \in \mathbb{R}^{n_x+1}$$

$$\hat{y} = \sigma(\Theta^T x)$$

$$\Theta = \begin{bmatrix} \theta_0 \\ \theta_1 \\ \theta_2 \\ \vdots \\ \theta_{n_x} \end{bmatrix} \begin{matrix} \} b \leftarrow \\ \\ \\ \\ \} w \leftarrow \end{matrix}$$

$$\sigma(z) = \frac{1}{1 + e^{-z}}$$

If z large $\sigma(z) \approx \frac{1}{1+0} = 1$

If z large negative number

$$\sigma(z) = \frac{1}{1 + e^{-z}} \approx \frac{1}{1 + \text{Big num}} \approx 0$$

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Basics of Neural Network Programming

Logistic Regression cost function

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Logistic Regression cost function

→ $\hat{y}^{(i)} = \sigma(w^T x^{(i)} + b)$, where $\sigma(z^{(i)}) = \frac{1}{1+e^{-z^{(i)}}}$ $z^{(i)} = w^T x^{(i)} + b$

Given $\{(x^{(1)}, y^{(1)}), \dots, (x^{(m)}, y^{(m)})\}$, want $\hat{y}^{(i)} \approx y^{(i)}$.

Loss (error) function:

$$\mathcal{L}(\hat{y}, y) = \frac{1}{2} (\hat{y} - y)^2$$

$$\mathcal{L}(\hat{y}, y) = - (y \log \hat{y} + (1-y) \log (1-\hat{y})) \leftarrow$$

If $y=1$: $\mathcal{L}(\hat{y}, y) = -\log \hat{y} \leftarrow$ want $\log \hat{y}$ large, want $\hat{y}$ large.

If $y=0$: $\mathcal{L}(\hat{y}, y) = -\log (1-\hat{y}) \leftarrow$ want $\log (1-\hat{y})$ large want $\hat{y}$ small

Cost function: $J(w, b) = \frac{1}{m} \sum_{i=1}^m \mathcal{L}(\hat{y}^{(i)}, y^{(i)}) = \frac{1}{m} \sum_{i=1}^m [y^{(i)} \log \hat{y}^{(i)} + (1-y^{(i)}) \log (1-\hat{y}^{(i)})]$

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Basics of Neural Network Programming

Gradient Descent

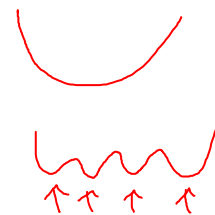
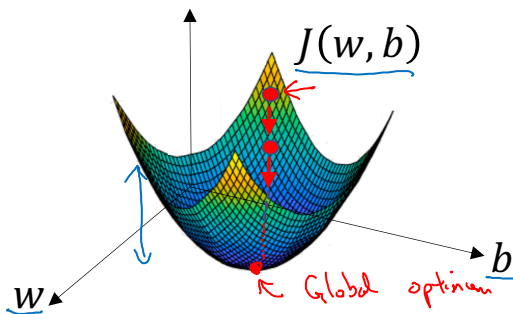
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Gradient Descent

Recap: $\hat{y} = \sigma(w^T x + b)$, $\sigma(z) = \frac{1}{1+e^{-z}}$ ←

$$J(w, b) = \frac{1}{m} \sum_{i=1}^m \mathcal{L}(\hat{y}^{(i)}, y^{(i)}) = -\frac{1}{m} \sum_{i=1}^m y^{(i)} \log \hat{y}^{(i)} + (1 - y^{(i)}) \log(1 - \hat{y}^{(i)})$$

Want to find w, b that minimize $J(w, b)$

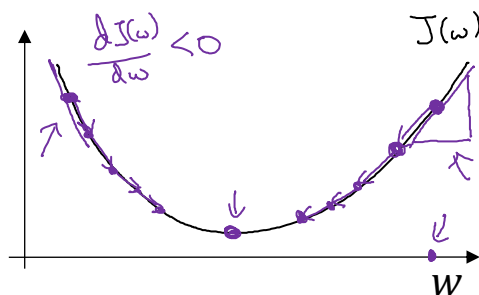


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Gradient Descent



Repeat {
 $w := w - \alpha \frac{dJ(w)}{dw}$
 }
 $w := w - \alpha dw$

learning rate α is indicated by a green arrow pointing to the α in the equation.

$$\frac{dJ(w)}{dw} = ?$$

$$J(w, b)$$

$$w := w - \alpha \frac{\partial J(w, b)}{\partial w}$$

$$b := b - \alpha \frac{\partial J(w, b)}{\partial b}$$

Handwritten notes include:

- $\frac{\partial J(w, b)}{\partial w}$ and $\frac{\partial J(w, b)}{\partial b}$ are boxed in red.
- Arrows point from these boxes to the terms dw and db respectively.
- A note "partial derivatin" (sic) points to the partial derivative symbols ∂ .

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Basics of Neural Network Programming

Logistic Regression Gradient descent

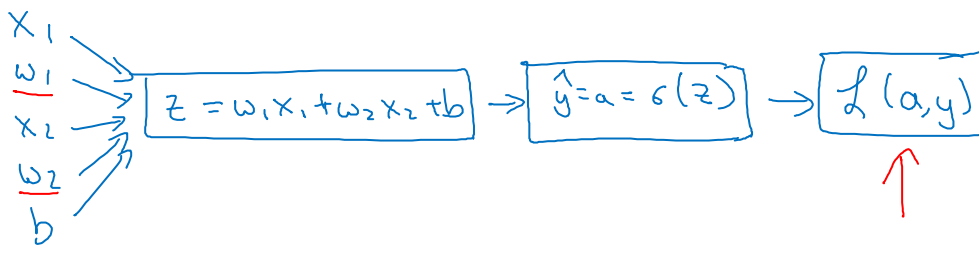
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Logistic regression recap

$$\rightarrow z = w^T x + b$$

$$\rightarrow \hat{y} = a = \sigma(\underline{z})$$

$$\rightarrow \mathcal{L}(a, y) = -(y \log(a) + (1 - y) \log(1 - a))$$

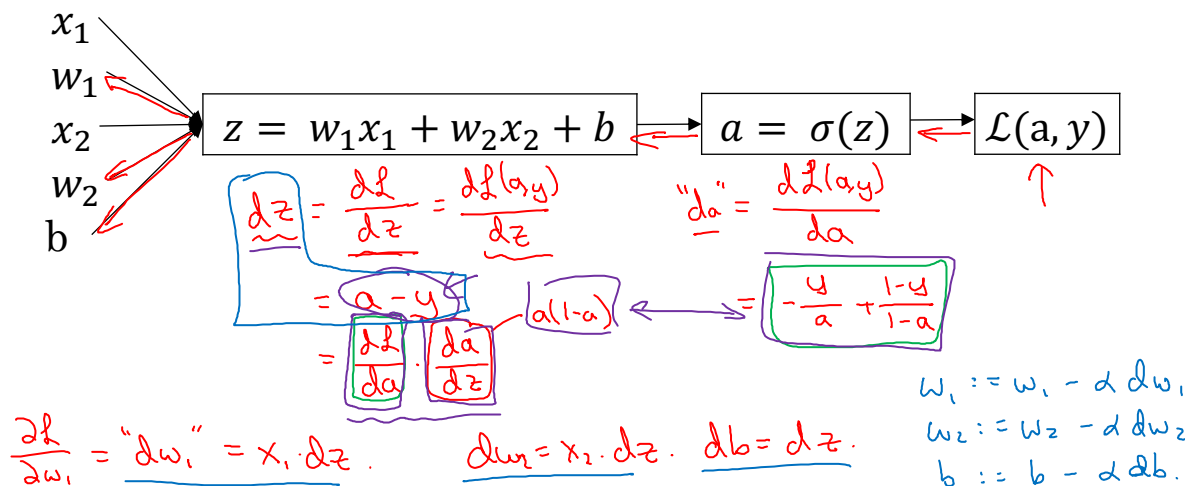


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Logistic regression derivatives



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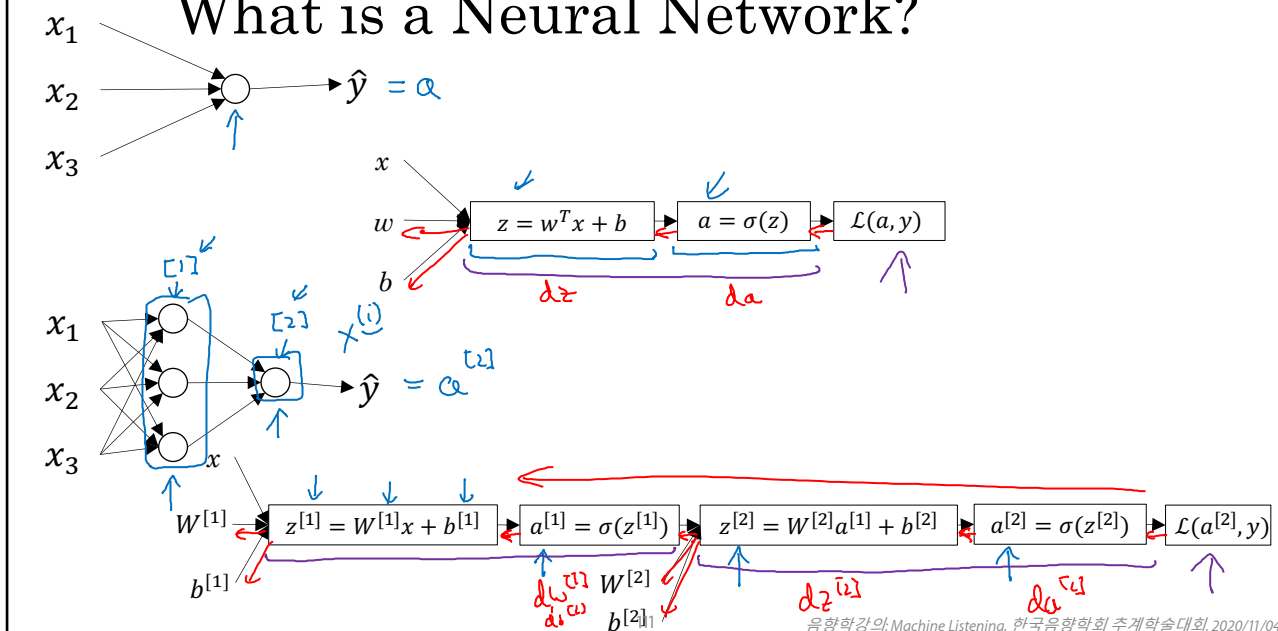
One hidden layer Neural Network

Neural Networks Overview

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What is a Neural Network?



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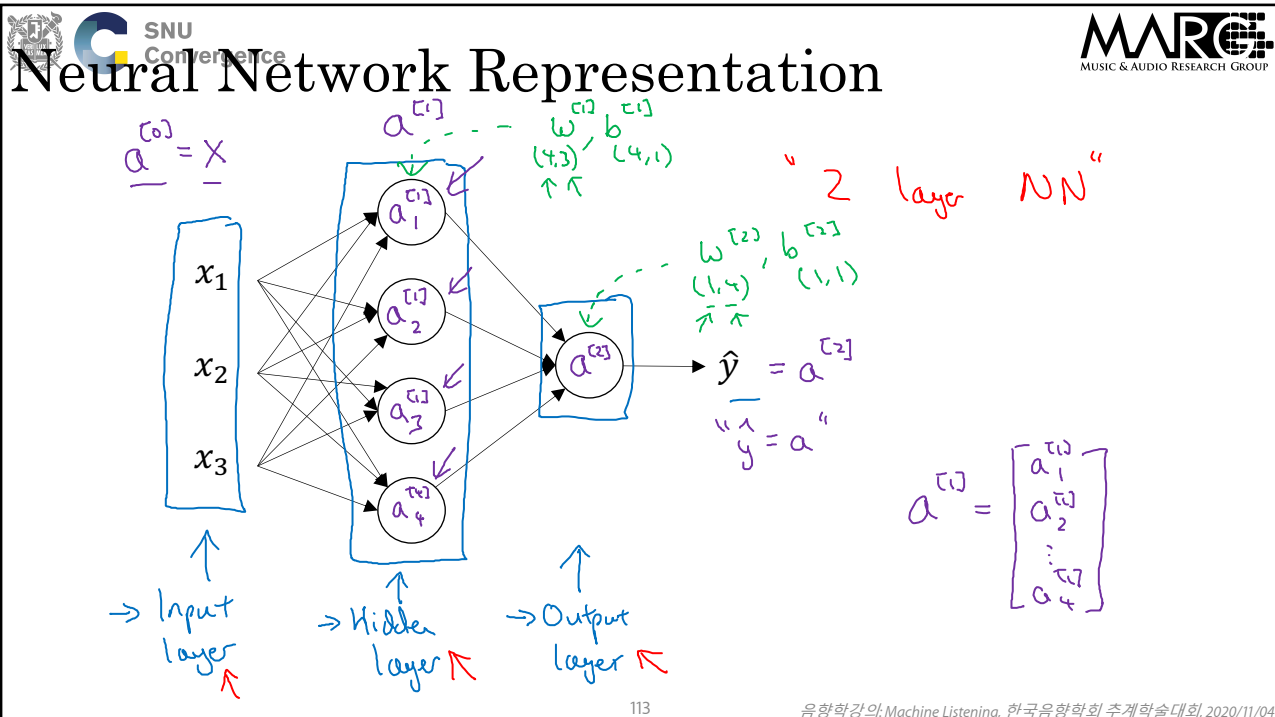


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One hidden layer Neural Network

Neural Network Representation

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One hidden layer Neural Network

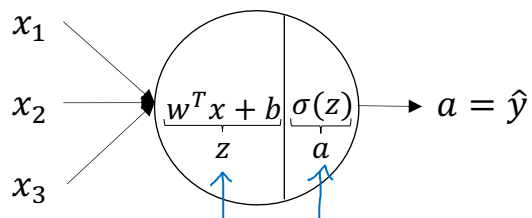
Computing a Neural Network's Output

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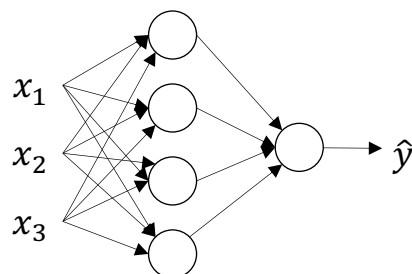
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Neural Network Representation



$$z = w^T x + b$$

$$a = \sigma(z)$$

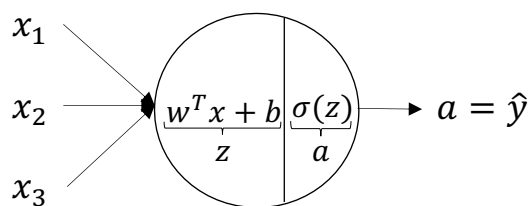


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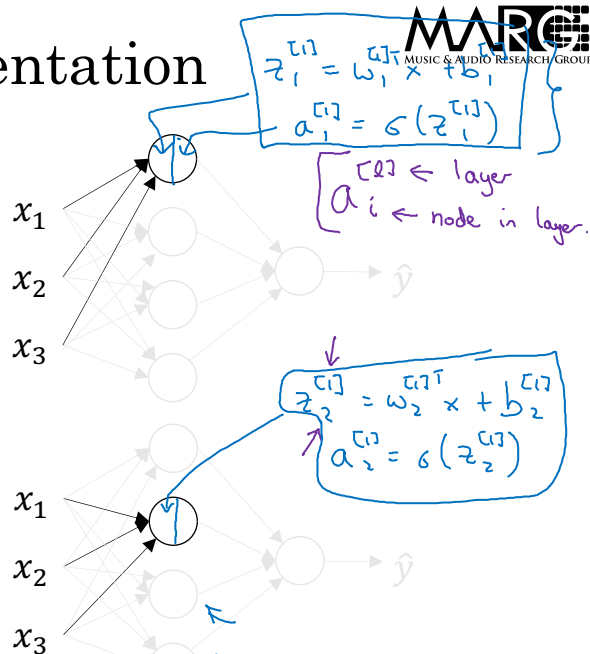
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Neural Network Representation



$$z = w^T x + b$$

$$a = \sigma(z)$$



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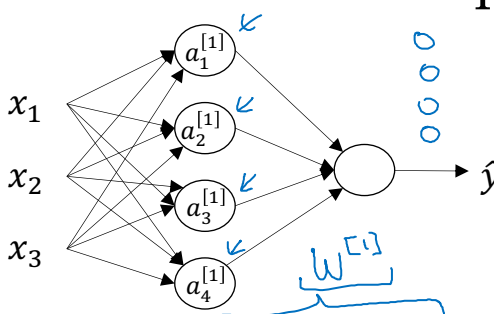
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Neural Network Representation



x_1, x_2, x_3

$a_1^{[1]}, a_2^{[1]}, a_3^{[1]}, a_4^{[1]}$

$\hat{y}$

$W^{[1]}$

$z_1^{[1]} = w_1^{[1]T} x + b_1^{[1]}$
 $z_2^{[1]} = w_2^{[1]T} x + b_2^{[1]}$
 $z_3^{[1]} = w_3^{[1]T} x + b_3^{[1]}$
 $z_4^{[1]} = w_4^{[1]T} x + b_4^{[1]}$

$a_1^{[1]} = \sigma(z_1^{[1]})$
 $a_2^{[1]} = \sigma(z_2^{[1]})$
 $a_3^{[1]} = \sigma(z_3^{[1]})$
 $a_4^{[1]} = \sigma(z_4^{[1]})$

$\Rightarrow z^{[1]} = \begin{bmatrix} w_1^{[1]T} \\ w_2^{[1]T} \\ w_3^{[1]T} \\ w_4^{[1]T} \end{bmatrix} \begin{bmatrix} x_1 \\ x_2 \\ x_3 \end{bmatrix} + \begin{bmatrix} b_1^{[1]} \\ b_2^{[1]} \\ b_3^{[1]} \\ b_4^{[1]} \end{bmatrix}$

$\Rightarrow a^{[1]} = \begin{bmatrix} a_1^{[1]} \\ a_2^{[1]} \\ a_3^{[1]} \\ a_4^{[1]} \end{bmatrix} = \sigma(z^{[1]})$

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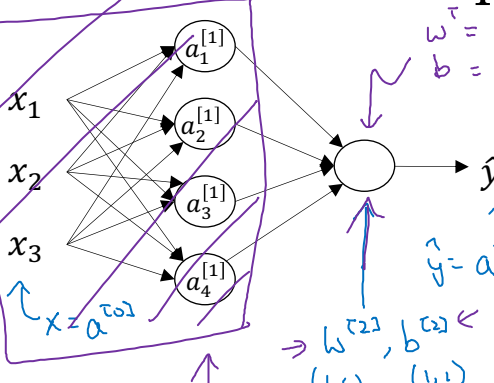
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Neural Network Representation learning



x_1, x_2, x_3

$a_1^{[1]}, a_2^{[1]}, a_3^{[1]}, a_4^{[1]}$

$\hat{y}$

$W^{[2]}, b^{[2]}$

$x = a^{[0]}$

$\hat{y} = a^{[2]}$

$z = w^T x + b$

$\hat{y} = a = \sigma(z)$

Given input x :

$\Rightarrow z^{[1]} = W^{[1]} a^{[0]} + b^{[1]}$
 $(4,1) \quad (4,3) \quad (3,1) \quad (4,1)$

$\Rightarrow a^{[1]} = \sigma(z^{[1]})$
 $(4,1) \quad (4,1)$

$\Rightarrow z^{[2]} = W^{[2]} a^{[1]} + b^{[2]}$
 $(1,1) \quad (1,4) \quad (4,1) \quad (1,1)$

$\Rightarrow a^{[2]} = \sigma(z^{[2]})$
 $(1,1) \quad (1,1)$

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One hidden layer Neural Network

Vectorizing across multiple examples

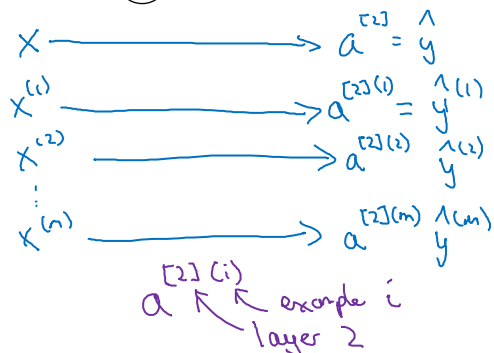
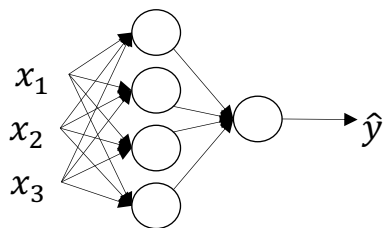
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Vectorizing across multiple examples



$$\left\{ \begin{array}{l} z^{[1]} = W^{[1]}x + b^{[1]} \\ a^{[1]} = \sigma(z^{[1]}) \\ z^{[2]} = W^{[2]}a^{[1]} + b^{[2]} \\ a^{[2]} = \sigma(z^{[2]}) \end{array} \right\} \leftarrow$$

→ for $i = 1$ to n ,

$$\begin{array}{l} z^{[1](i)} = W^{[1]} x^{(i)} + b^{[1]} \\ a^{[1](i)} = \sigma(z^{[1](i)}) \\ z^{[2](i)} = W^{[2]} a^{[1](i)} + b^{[2]} \\ a^{[2](i)} = \sigma(z^{[2](i)}) \end{array}$$

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Vectorizing across multiple examples

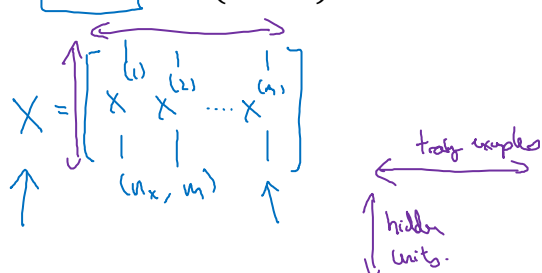
for $i = 1$ to m :

$$z^{[1]}(i) = W^{[1]}x^{(i)} + b^{[1]}$$

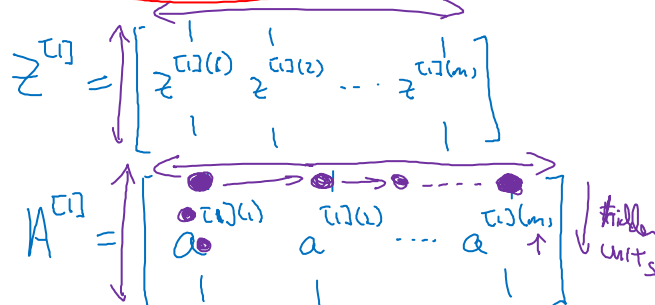
$$a^{[1]}(i) = \sigma(z^{[1]}(i))$$

$$z^{[2]}(i) = W^{[2]}a^{[1]}(i) + b^{[2]}$$

$$a^{[2]}(i) = \sigma(z^{[2]}(i))$$



$$\begin{aligned} z^{[1]} &= W^{[1]}X + b^{[1]} \\ \rightarrow A^{[1]} &= \sigma(z^{[1]}) \\ \rightarrow z^{[2]} &= W^{[2]}A^{[1]} + b^{[2]} \\ \rightarrow A^{[2]} &= \sigma(z^{[2]}) \end{aligned}$$



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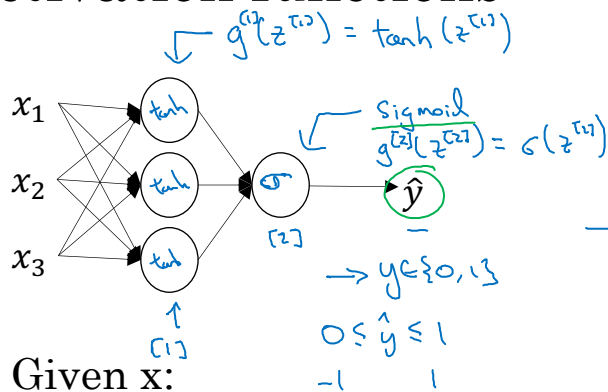
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One hidden layer Neural Network

Activation functions

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Activation functions



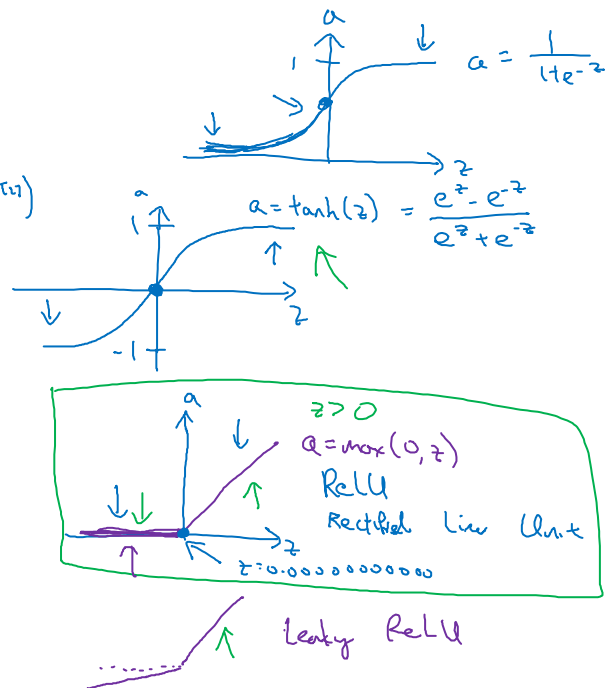
Given x :

$$z^{[1]} = W^{[1]}x + b^{[1]}$$

$$\rightarrow a^{[1]} = \sigma(z^{[1]}) \quad g(z^{(1)})$$

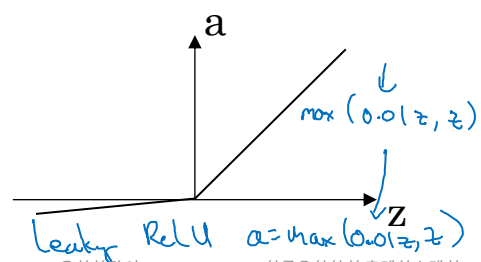
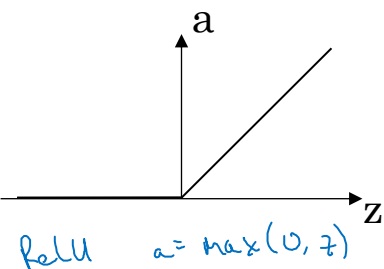
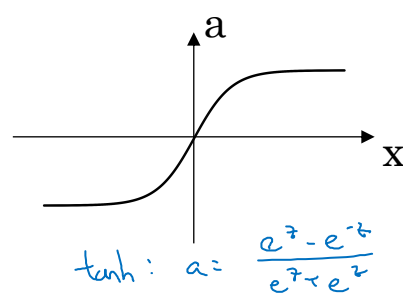
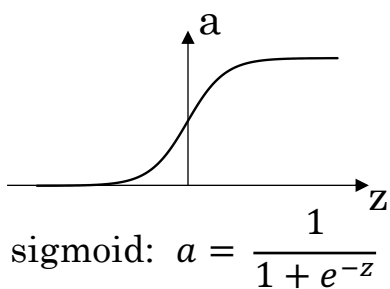
$$z^{[2]} = W^{[2]}a^{[1]} + b^{[2]}$$

$$\rightarrow a^{[2]} = \sigma(z^{[2]}) \quad g(z^{(2)})$$



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Pros and cons of activation functions



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One hidden layer Neural Network

Why do you need non-linear activation functions?

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Activation function

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Given x :

- $\rightarrow z^{[1]} = W^{[1]}x + b^{[1]}$
- $\rightarrow a^{[1]} = g^{[1]}(z^{[1]}) \geq z^{[1]}$
- $\rightarrow z^{[2]} = W^{[2]}a^{[1]} + b^{[2]}$
- $\rightarrow a^{[2]} = g^{[2]}(z^{[2]}) \geq z^{[2]}$

$g(z) = z$
"linear activation function"

$$a^{[1]} = z^{[1]} = W^{[1]}x + b^{[1]}$$

$$a^{[2]} = z^{[2]} = W^{[2]}a^{[1]} + b^{[2]}$$

$$a^{[2]} = W^{[2]}(W^{[1]}x + b^{[1]}) + b^{[2]}$$

$$= \underbrace{(W^{[2]}W^{[1]})}_w x + \underbrace{(W^{[2]}b^{[1]} + b^{[2]})}_{b'}$$

$$= w'x + b'$$

$$g(z) = z$$

$y \in \mathbb{R}$
\$0 \dots \\$1,000,000\$

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Derivatives of activation functions

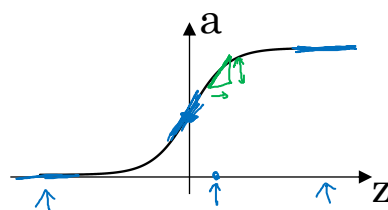
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Sigmoid activation function



$$g(z) = \frac{1}{1 + e^{-z}}$$

$$a = g(z) = \frac{1}{1 + e^{-z}}$$

$$g'(z) = \frac{d}{dz} g(z) = \text{slope of } g(z) \text{ at } z$$

$$= \frac{1}{1 + e^{-z}} \left(1 - \frac{1}{1 + e^{-z}} \right)$$

$$= g(z) (1 - g(z)) \leftarrow$$

$$= \boxed{a(1-a)} \quad \left| \begin{array}{l} g'(z) = a(1-a) \\ \uparrow \\ a \end{array} \right.$$

$$z = 10, \quad g(z) \approx 1$$

$$\frac{d}{dz} g(z) \approx 1(1-1) \approx 0$$

$$z = -10, \quad g(z) \approx 0$$

$$\frac{d}{dz} g(z) \approx 0(1-0) \approx 0$$

$$z = 0, \quad g(z) = \frac{1}{2}$$

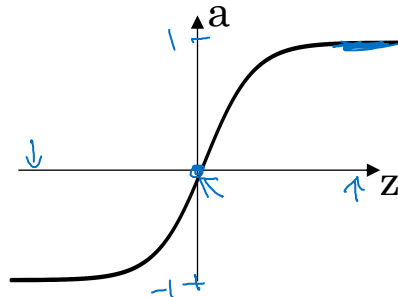
$$\frac{d}{dz} g(z) = \frac{1}{2} \left(1 - \frac{1}{2} \right) = \frac{1}{4}$$

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Tanh activation function



$$g(z) = \tanh(z) = \frac{e^z - e^{-z}}{e^z + e^{-z}}$$

$$g'(z) = \frac{d}{dz} g(z) = \text{slope of } g(z) \text{ at } z$$

$$= 1 - \tanh(z)^2 \leftarrow$$

$$a = g(z), \quad g'(z) = 1 - \underset{\uparrow}{a^2}$$

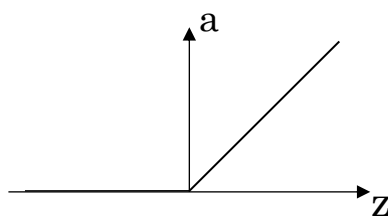
$$\begin{aligned} z=10 \quad \tanh(z) &\approx 1 \\ q'(z) &\approx 0 \\ z=-10 \quad \tanh(z) &\approx -1 \\ q'(z) &\approx 0 \\ z=0 \quad \tanh(z) &= 0 \\ q'(z) &= 1 \end{aligned}$$

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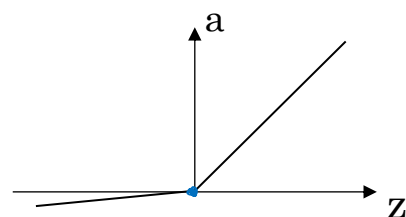
ReLU and Leaky ReLU



ReLU

$$g(z) = \max(0, z)$$

$$\rightarrow g'(z) = \begin{cases} 0 & \text{if } z < 0 \\ 1 & \text{if } z \geq 0 \end{cases}$$



Leaky ReLU

$$g(z) = \max(0.01z, z)$$

$$g'(z) = \begin{cases} 0.01 & \text{if } z < 0 \\ 1 & \text{if } z \geq 0 \end{cases}$$

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One hidden layer Neural Network

Gradient descent for neural networks

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Gradient descent for neural networks

Parameters: $w^{[1]}, b^{[1]}, w^{[2]}, b^{[2]}$
 $(n^{[1]}, n^{[2]})$ $(n^{[1]}, 1)$ $(n^{[2]}, n^{[3]})$ $(n^{[2]}, 1)$ $n_x = n^{[0]}, n^{[1]}, \underline{n^{[2]} = 1}$

Cost function: $J(w^{[1]}, b^{[1]}, \underline{w^{[2]}}, \underline{b^{[2]}}) = \frac{1}{m} \sum_{i=1}^m \ell(\hat{y}_i, y_i)$
 $\uparrow \quad \uparrow \quad \uparrow a^{[2]}$

Gradient descent:

→ Repeat {
 → Compute predicts $(\hat{y}^{(i)}, i=1, \dots, m)$
 $\underline{\frac{\partial J}{\partial w^{[2]}}} = \frac{\partial J}{\partial w^{[2]}} , \underline{\frac{\partial J}{\partial b^{[2]}}} = \frac{\partial J}{\partial b^{[2]}} , \dots$
 $w^{[2]} := w^{[2]} - \alpha \frac{\partial J}{\partial w^{[2]}}$
 $b^{[2]} := b^{[2]} - \alpha \frac{\partial J}{\partial b^{[2]}}$
 $w^{[2]} := \dots \quad b^{[2]} := \dots$

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Formulas for computing derivatives

Forward propagation:

$$\begin{aligned} z^{[1]} &= W^{[1]}x + b^{[1]} \\ A^{[1]} &= g^{[1]}(z^{[1]}) \leftarrow \\ z^{[2]} &= W^{[2]}A^{[1]} + b^{[2]} \\ A^{[2]} &= g^{[2]}(z^{[2]}) = \sigma(z^{[2]}) \end{aligned}$$

Back propagation:

$$\begin{aligned} \delta z^{[2]} &= A^{[2]} - Y \quad Y = [y^{(1)} \ y^{(2)} \dots \ y^{(m)}] \\ \delta W^{[2]} &= \frac{1}{m} \delta z^{[2]} A^{[1]T} \\ \delta b^{[2]} &= \frac{1}{m} \text{np.sum}(\delta z^{[2]}, \text{axis}=1, \text{keepdims}=\text{True}) \\ \delta z^{[1]} &= \underbrace{W^{[2]T}}_{(n^{[2]}, m)} \delta z^{[2]} \star \underbrace{g^{[1]'}(z^{[1]})}_{\text{element-wise product}} \quad (n^{[1]}, m) \\ \delta W^{[1]} &= \frac{1}{m} \delta z^{[1]} x^T \\ \delta b^{[1]} &= \frac{1}{m} \text{np.sum}(\delta z^{[1]}, \text{axis}=1, \text{keepdims}=\text{True}) \end{aligned}$$

$(n^{[2]}, 1)$ $(n^{[1]}, 1)$ $\text{reshape} \uparrow$

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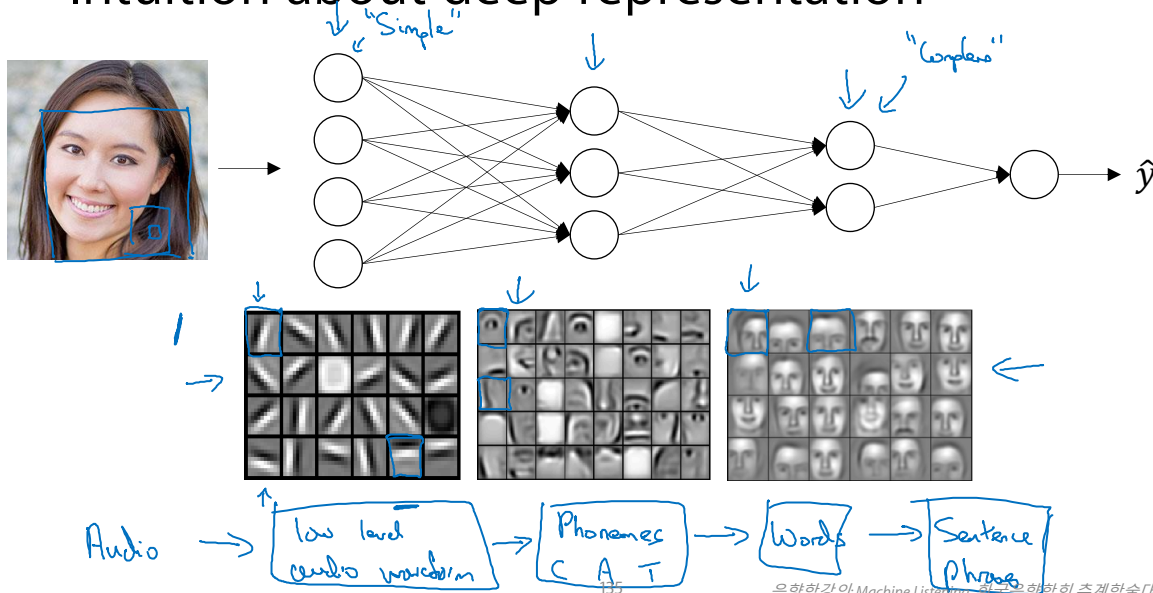
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Deep Neural Networks

Why deep representations?

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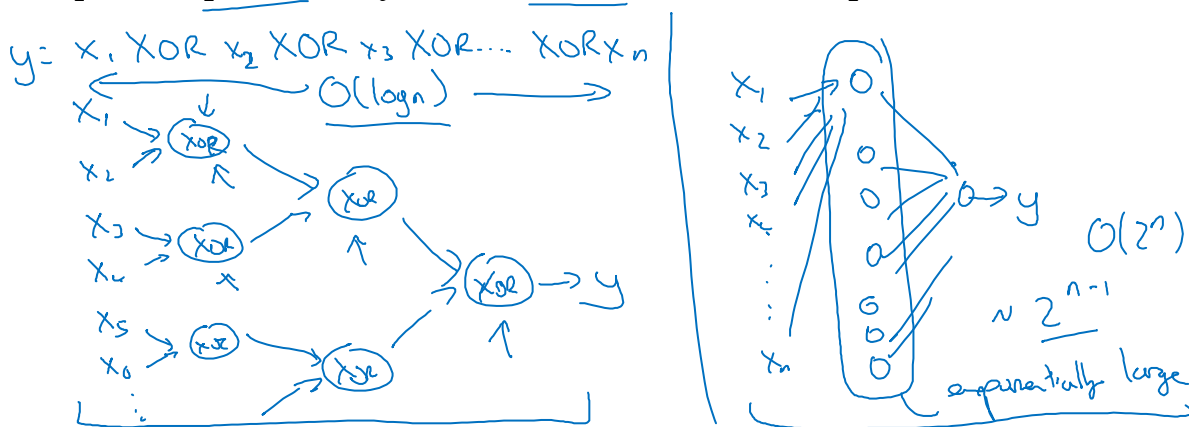
Intuition about deep representation



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Circuit theory and deep learning

Informally: There are functions you can compute with a “small” L-layer deep neural network that shallower networks require exponentially more hidden units to compute.



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Deep Neural Networks

Building blocks of deep neural networks

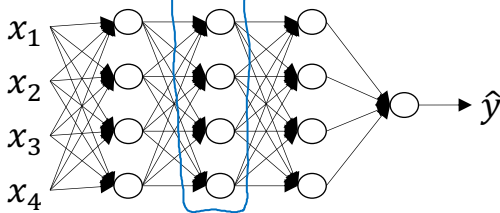
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Forward and backward functions



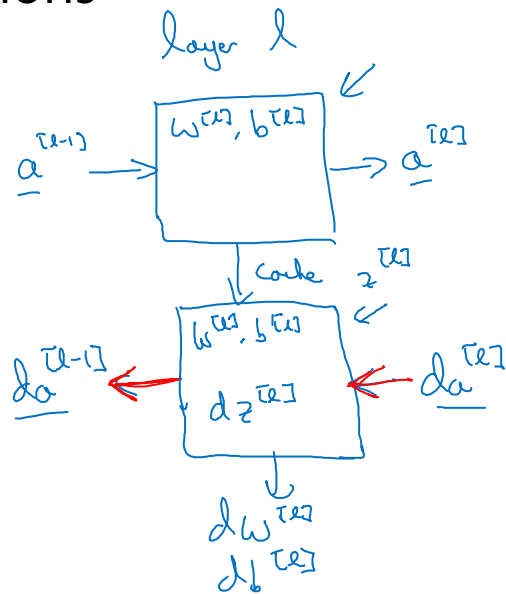
layer l : $W^{[l]}, b^{[l]}$

→ Forward: Input $a^{[l-1]}$, output $a^{[l]}$

$$z^{[l]} = W^{[l]} a^{[l-1]} + b^{[l]} \quad \text{cache } z^{[l]}$$

$$a^{[l]} = g(z^{[l]})$$

→ Backward: Input $da^{[l]}$, output $da^{[l-1]}$
cache $(z^{[l]})$
 $\frac{dL}{dz^{[l]}}$
 $\frac{dL}{db^{[l]}}$

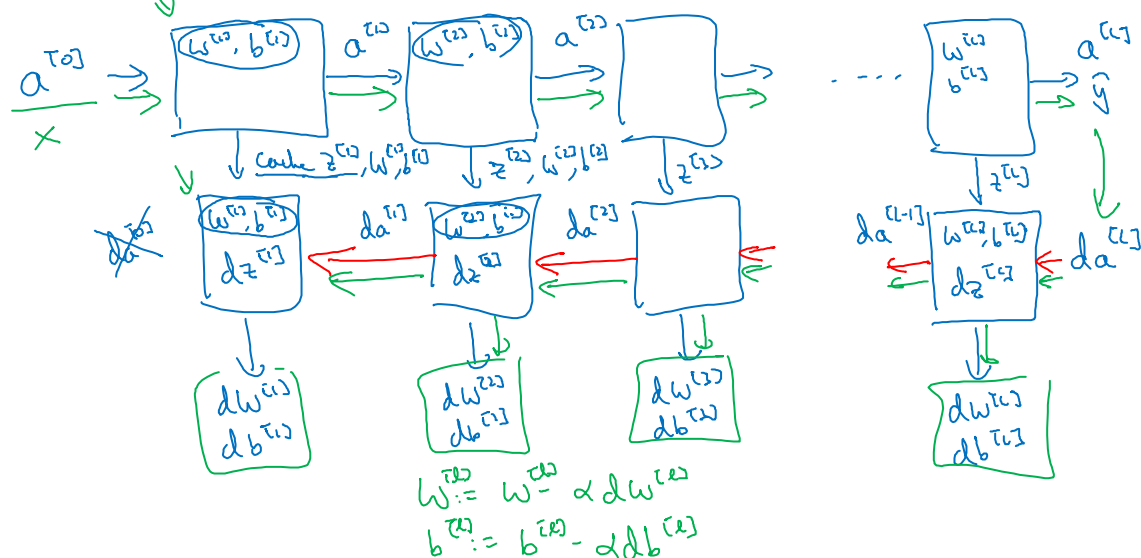


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Forward and backward functions



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Deep Neural Networks

Parameters vs Hyperparameters

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What are hyperparameters?

Parameters: $W^{[1]}, b^{[1]}, W^{[2]}, b^{[2]}, W^{[3]}, b^{[3]} \dots$

Hyperparameters: α , $\frac{\alpha}{\tau}$, #iterations, #hidden layers L , #hidden units $n^{[1]}, n^{[2]}, \dots$, choice of activation function

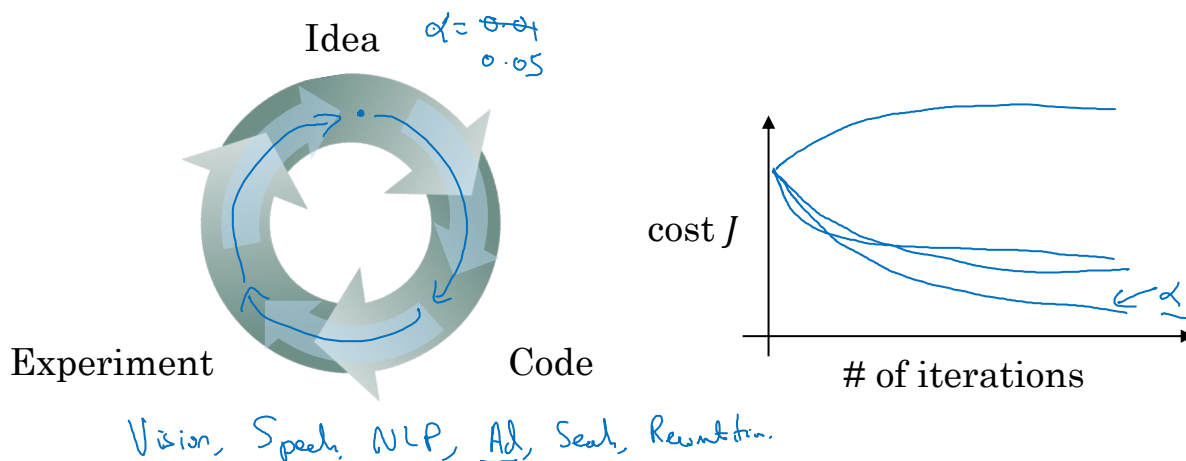
Loss: Momentum, mini-batch size, regularizations, ...

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Applied deep learning is a very empirical process



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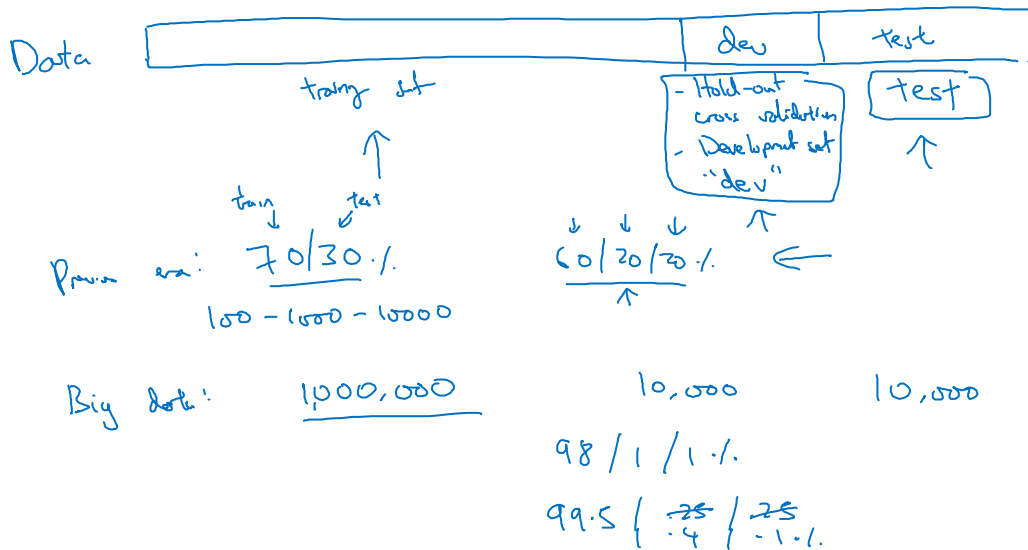
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Setting up your ML application

Train/dev/test sets

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Train/dev/test sets



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Mismatched train/test distribution

Training set:

Cat pictures from
webpages

Dev/test sets:

Cat pictures from
users using your app

→ Make sure dev and test come from same distribution.

↓ ↓
train / dev "test"

↓ ↓
train / dev train / test

→ train / dev

Not having a test set might be okay. (Only dev set.)

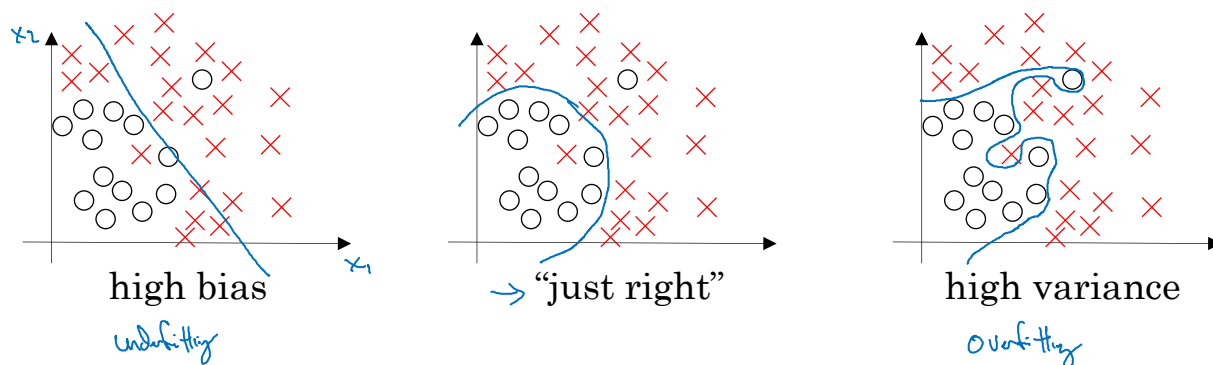


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Setting up your ML application

Bias/Variance

Bias and Variance



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Bias and Variance Cat classification



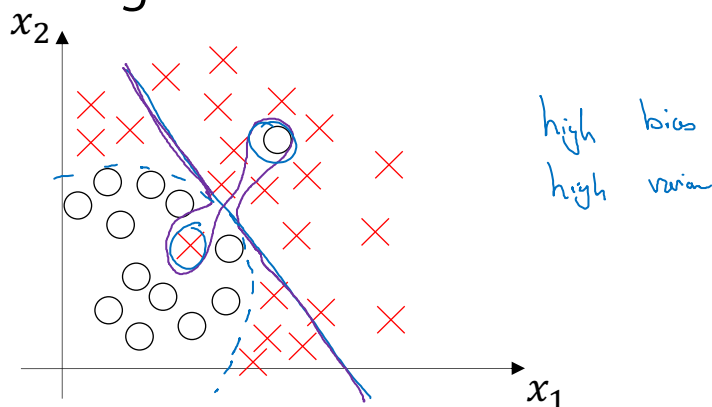
		$y=1$	$y=0$
			
Train set error:	1%	15% $\swarrow$	15%
Dev set error:	11%	16% $\swarrow$	0.5%
	high variance $\uparrow$	high bias $\uparrow \uparrow$	low bias $\uparrow$
		high bias & high variance	low variance $\uparrow$
Human: 20%			
Optimal (Bayes) error: 2% to 15%			
		Blurry images	

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High bias and high variance



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Regularizing your neural network

Regularization

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Logistic regression

$$\min_{w,b} J(w,b)$$

$$w \in \mathbb{R}^{n_x}, b \in \mathbb{R}$$

λ = regularization parameter
lambda

$$J(w,b) = \frac{1}{m} \sum_{i=1}^m \ell(y^{(i)}, \hat{y}^{(i)}) + \frac{\lambda}{2m} \|w\|_2^2$$

$$+ \frac{\lambda}{2m} b^2$$

omit

L_2 regularization $\|w\|_2^2 = \sum_{j=1}^{n_x} w_j^2 = w^T w \leftarrow$

L_1 regularization $\frac{\lambda}{2m} \sum_{j=1}^{n_x} |w_j| = \frac{\lambda}{2m} \|w\|_1$

w will be sparse

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Neural network

$$\rightarrow J(w^{[1]}, b^{[1]}, \dots, w^{[L]}, b^{[L]}) = \frac{1}{m} \sum_{i=1}^m \ell(y^{(i)}, \hat{y}^{(i)}) + \frac{\lambda}{2m} \sum_{l=1}^L \|w^{[l]}\|_F^2$$

$$\|w^{[l]}\|_F^2 = \sum_{i=1}^{n^{[l]}} \sum_{j=1}^{n^{[l-1]}} (w_{ij}^{[l]})^2$$

$w^{[l]}: (n^{[l]}, n^{[l-1]})$

"Frobenius norm" $\| \cdot \|_2^2$ $\| \cdot \|_F^2$

$$dw^{[l]} = (\text{from backprop}) + \frac{\lambda}{m} w^{[l]}$$

$$\frac{\partial J}{\partial w^{[l]}} = dw^{[l]}$$

$$\rightarrow w^{[l]} := w^{[l]} - \alpha \underbrace{dw^{[l]}}$$

"Weight decay"

$$w^{[l]} := w^{[l]} - \alpha \left[(\text{from backprop}) + \frac{\lambda}{m} w^{[l]} \right]$$

$$= w^{[l]} - \frac{\alpha \lambda}{m} w^{[l]} - \alpha (\text{from backprop})$$

$$= \underbrace{\left(1 - \frac{\alpha \lambda}{m}\right)}_{\leq 1} w^{[l]} - \alpha (\text{from backprop})$$

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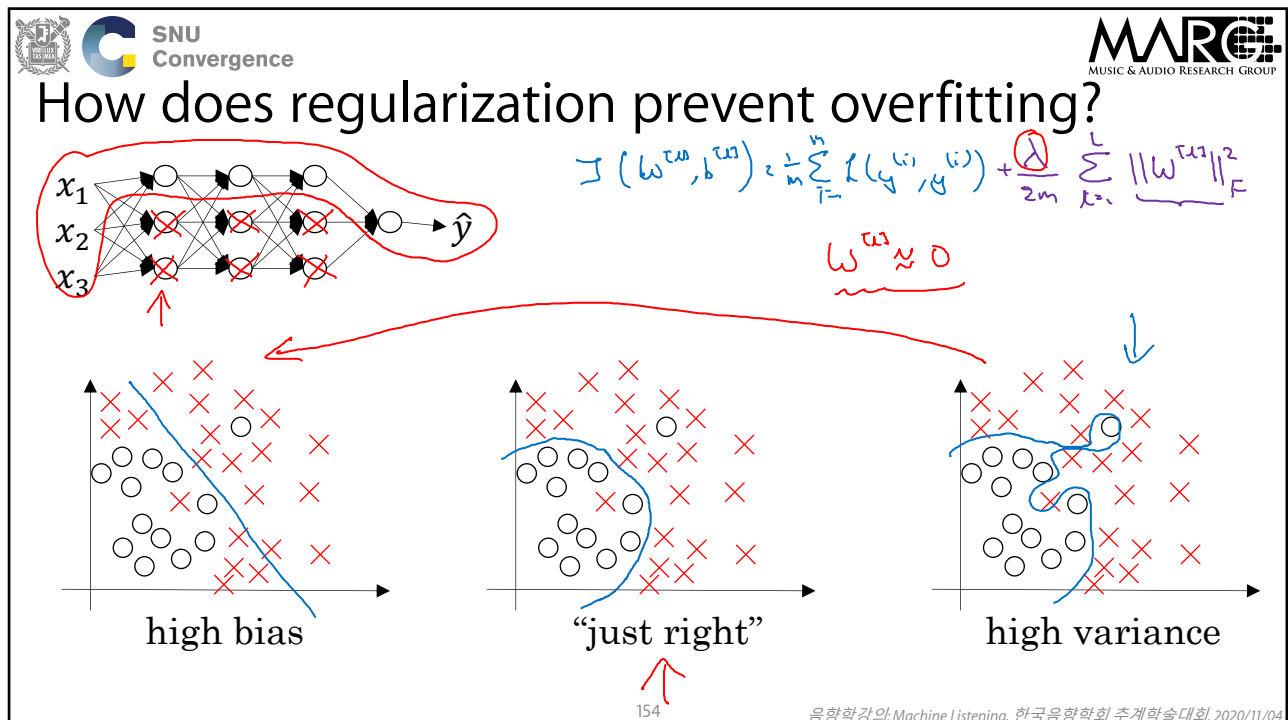


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Regularizing your neural network

Why regularization reduces overfitting

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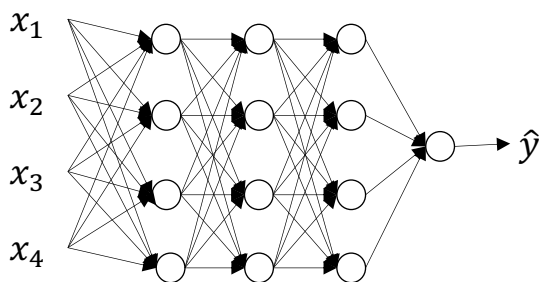
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Regularizing your neural network

Dropout regularization

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Dropout regularization



↑ ↑ ↑
0.5 0.5 0.5

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Implementing dropout ("Inverted dropout")

Illustrate with layer $l=3$. $\text{keep-prob} = \frac{0.8}{x}$ $\frac{0.2}{x}$

$\rightarrow d3 = \text{np.random.rand}(a3.\text{shape}[0], a3.\text{shape}[1]) < \text{keep-prob}$

$a3 = \text{np.multiply}(a3, d3)$ $\# a3 \times = d3$

$\rightarrow a3 /= \text{keep-prob}$

50 units. $\rightarrow$ 10 units shut off

$$z^{[4]} = w^{[4]} \cdot \underbrace{a^{[3]}}_{\substack{\uparrow \\ \text{reduced by } 20\% \\ /= 0.8}} + b^{[4]}$$

Test

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Making predictions at test time

$a^{[0]} = x$

No drop out.

$$z^{[1]} = w^{[1]} a^{[0]} + b^{[1]}$$

$$a^{[1]} = g^{[1]}(z^{[1]})$$

$$z^{[2]} = w^{[2]} a^{[1]} + b^{[2]}$$

$$a^{[2]} = \dots$$

$\downarrow$
 $\hat{y}$

$/= \text{keep-prob}$

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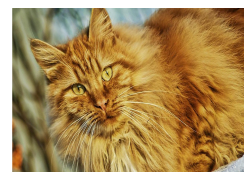
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Regularizing your neural network

Other regularization methods

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Data augmentation



4



4



4

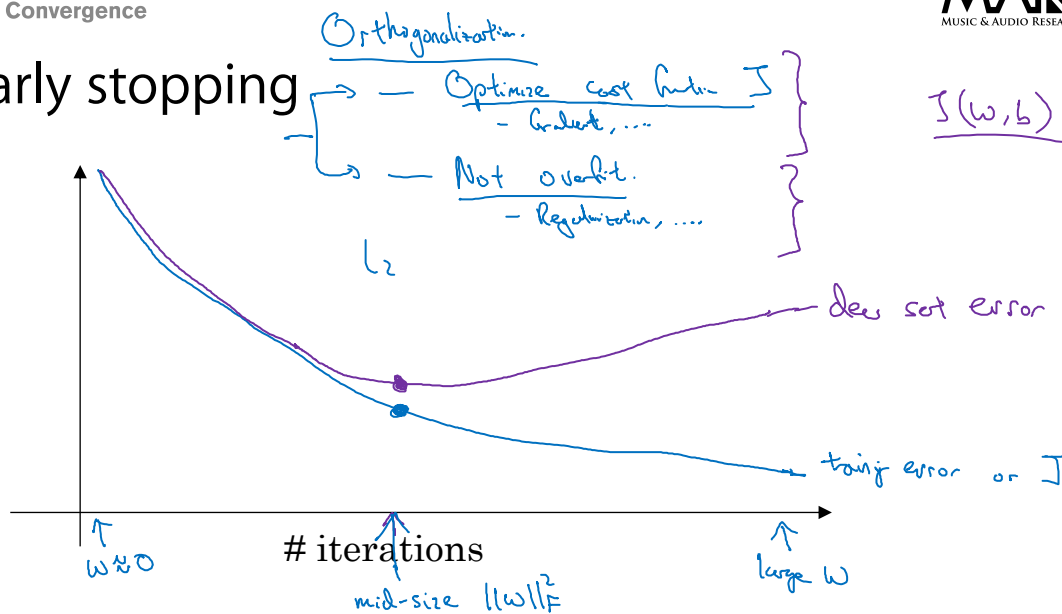


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Early stopping



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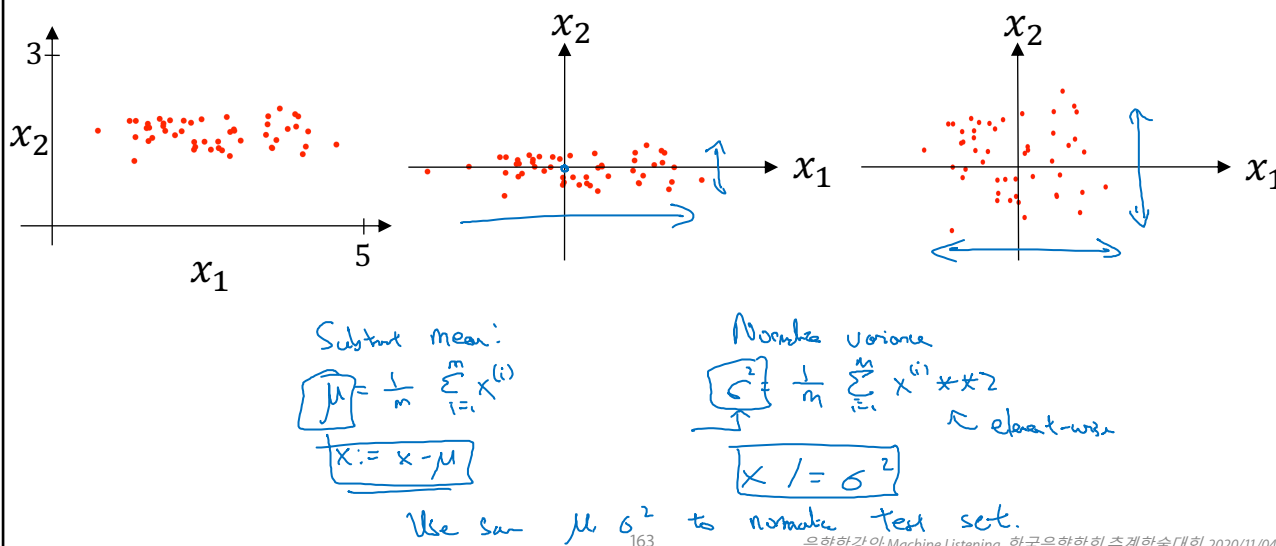
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Setting up your optimization problem

Normalizing inputs

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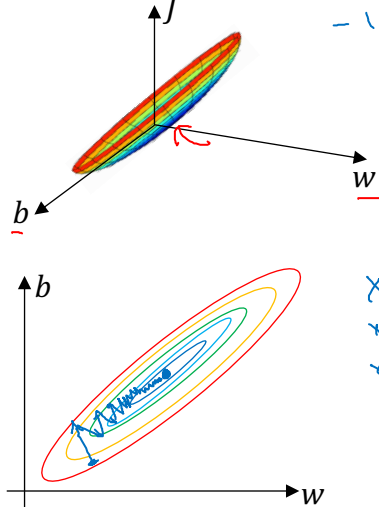
Normalizing training sets



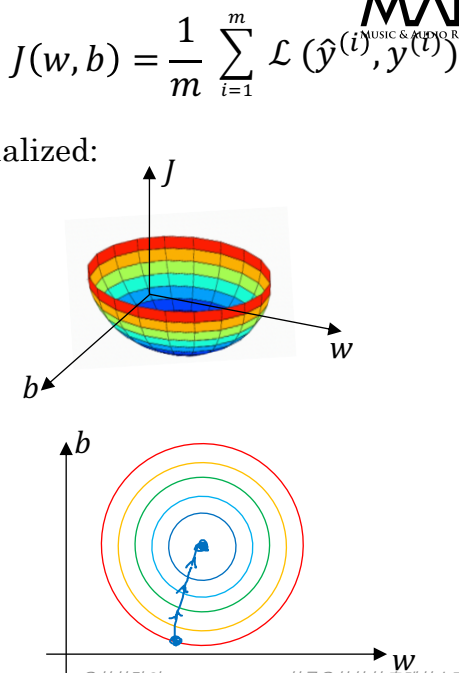
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Why normalize inputs?

Unnormalized:



Normalized:



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Optimization Algorithms

Mini-batch gradient descent

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Batch vs. mini-batch gradient descent

Vectorization allows you to efficiently compute on m examples.

$$X = \begin{bmatrix} x^{(1)} & x^{(2)} & x^{(3)} & \dots & x^{(1000)} & | & x^{(1001)} & \dots & x^{(2000)} & | & \dots & | & \dots & x^{(m)} \end{bmatrix}$$

(n_x, m) $X^{\{1\}} (n_x, 1000)$ $X^{\{2\}} (n_x, 1000)$ $X^{\{5,000\}} (n_x, 1000)$

$$Y = \begin{bmatrix} y^{(1)} & y^{(2)} & y^{(3)} & \dots & y^{(1000)} & | & y^{(1001)} & \dots & y^{(2000)} & | & \dots & | & \dots & y^{(m)} \end{bmatrix}$$

$(1, m)$ $Y^{\{1\}} (1, 1000)$ $Y^{\{2\}} (1, 1000)$ $Y^{\{5,000\}} (1, 1000)$

What if $m = 5,000,000$?

5,000 mini-batches of 1,000 each

Mini-batch t : $X^{\{t\}}, Y^{\{t\}}$

$x^{(i)}$

$z^{[l]}$

$X^{\{t\}}, Y^{\{t\}}$

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Mini-batch gradient descent

repeat {
 for $t = 1, \dots, 5000$ {
 Forward prop on X^{set} .
 $Z^{(t)} = W^{(t)} X^{set} + b^{(t)}$
 $A^{(t)} = g^{(t)}(Z^{(t)})$
 $\vdots$
 $A^{(t)} = g^{(t)}(Z^{(t)})$ } Vertical implentation (1000 example)
 Compute cost $J^{set} = \frac{1}{1000} \sum_{i=1}^L \mathcal{L}(\hat{y}^{(i)}, y^{(i)}) + \frac{\lambda}{2 \cdot 1000} \sum \|W^{(t)}\|^2$.
 Backprop to compute gradients w.r.t J^{set} (using (X^{set}, Y^{set}))
 $W^{(t+1)} = W^{(t)} - \alpha \Delta W^{(t)}$, $b^{(t+1)} = b^{(t)} - \alpha \Delta b^{(t)}$
 }
 } "1 epoch" pass through training set.

1 step of epoch using X^{set}, Y^{set} . (as if $m=1000$)

X, Y

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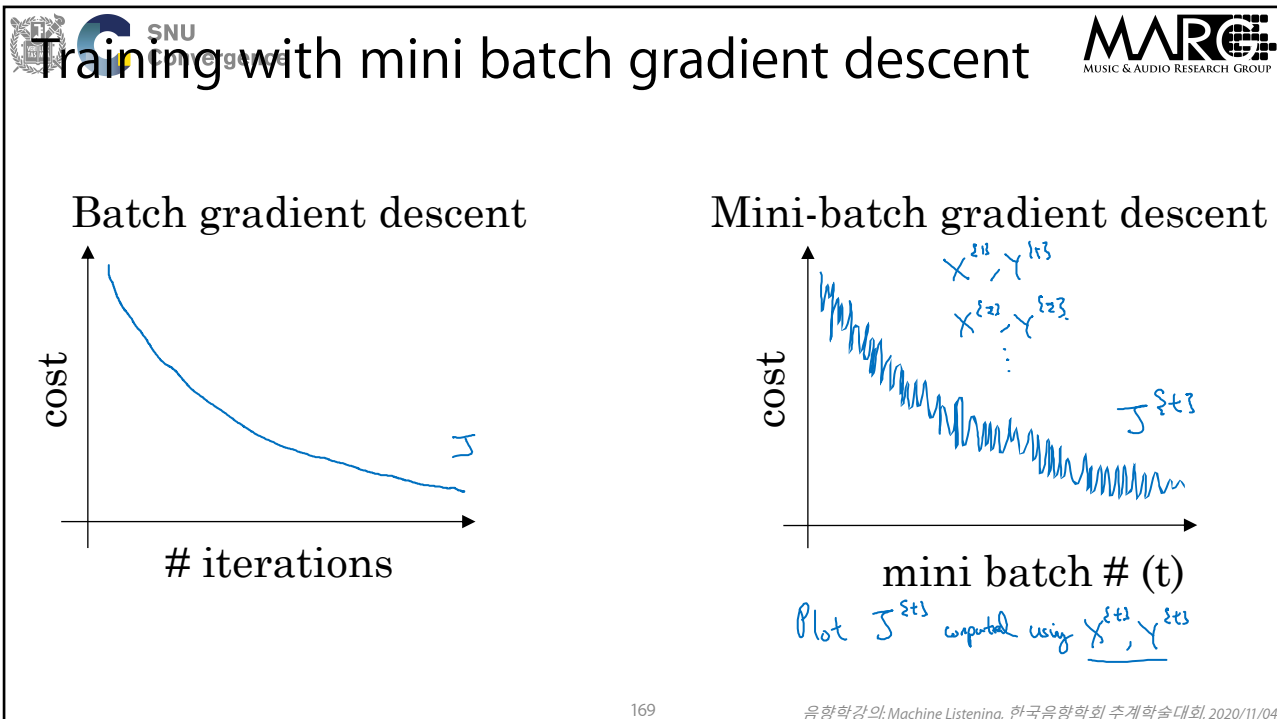
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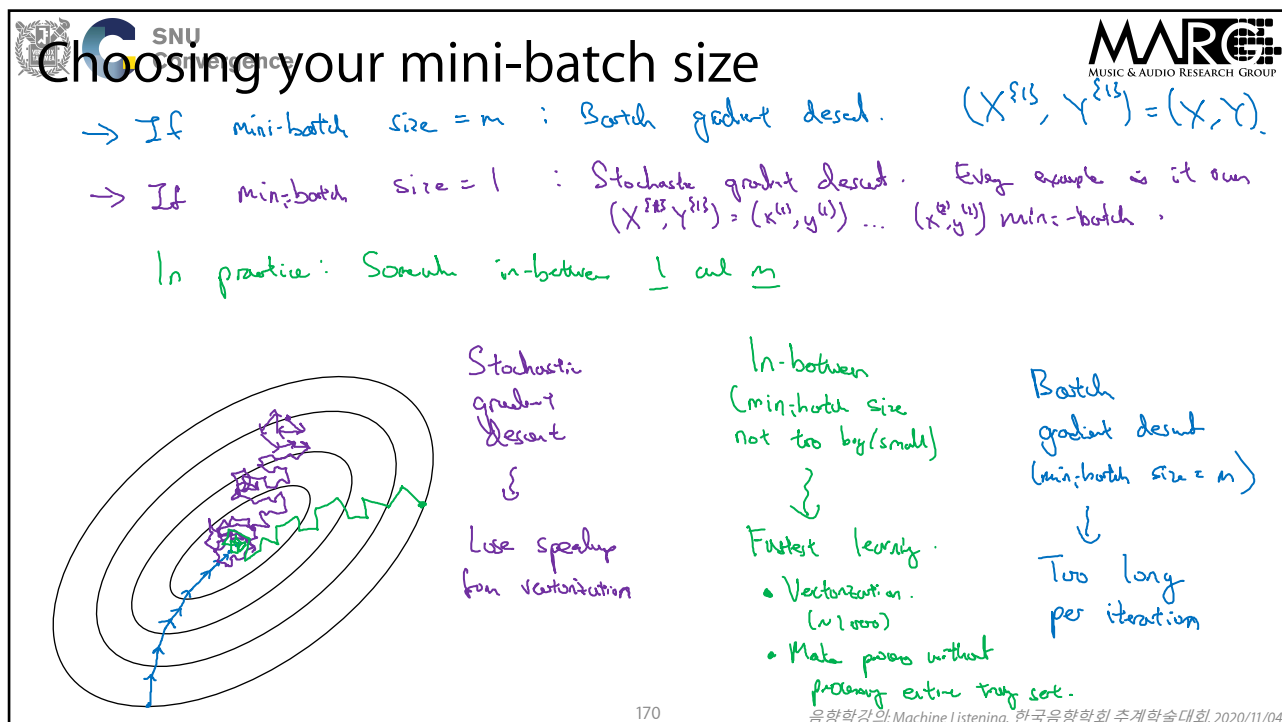
Optimization Algorithms

Understanding mini-batch gradient descent

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Choosing your mini-batch size

If small toy set : Use batch gradient descent.
($m \leq 2000$)

Typical mini-batch sizes:

→ 64, 128, 256, 512, 1024
 $2^6, 2^7, 2^8, 2^9, 2^{10}$

Make sure mini-batch fit in CPU/GPU memory.
 $X^{(k)}, y^{(k)}$



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Optimization Algorithms

Exponentially weighted averages

Temperature in London

$$\theta_1 = 40^\circ\text{F} \quad 4^\circ\text{C} \leftarrow$$

$$\theta_2 = 49^\circ\text{F} \quad 9^\circ\text{C}$$

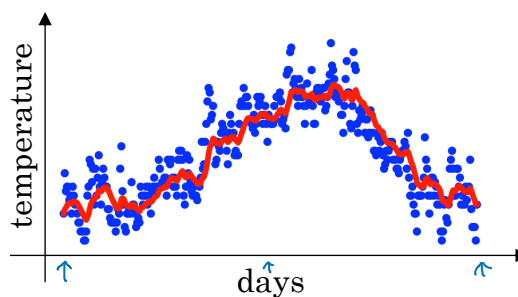
$$\theta_3 = 45^\circ\text{F} \quad \vdots$$

$\vdots$

$$\theta_{180} = 60^\circ\text{F} \quad 15^\circ\text{C}$$

$$\theta_{181} = 56^\circ\text{F} \quad \vdots$$

$\vdots$



$$\begin{aligned} V_0 &= 0 \\ V_1 &= 0.9 V_0 + 0.1 \theta_1 \\ V_2 &= 0.9 V_1 + 0.1 \theta_2 \\ V_3 &= 0.9 V_2 + 0.1 \theta_3 \\ &\vdots \\ V_t &= 0.9 V_{t-1} + 0.1 \theta_t \end{aligned}$$

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Exponentially weighted averages

$$V_t = \beta V_{t-1} + (1-\beta) \theta_t \leftarrow$$

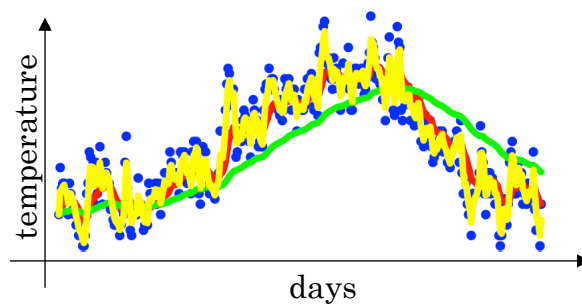
$\beta = 0.9$: ≈ 10 days' temper.

$\beta = 0.98$: ≈ 50 days

$\beta = 0.5$: ≈ 2 days

V_t is approximately
average over
 $\rightarrow \approx \frac{1}{1-\beta}$ days'
temperature.

$$\frac{1}{1-0.98} = 50$$



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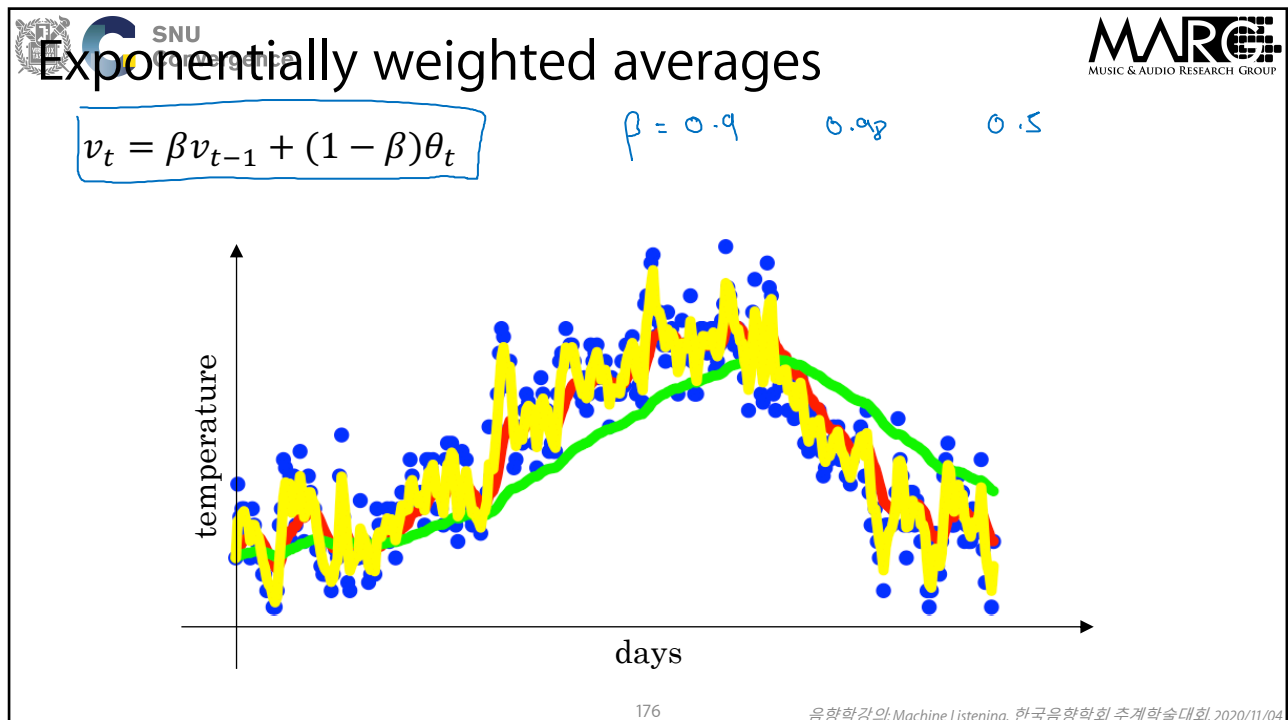


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Optimization Algorithms

Understanding exponentially weighted averages

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Exponentially weighted averages

$$v_t = \beta v_{t-1} + (1 - \beta) \theta_t$$

$$v_{100} = 0.9v_{99} + 0.1\theta_{100}$$

$$v_{99} = 0.9v_{98} + 0.1\theta_{99}$$

$$v_{98} = 0.9v_{97} + 0.1\theta_{98}$$

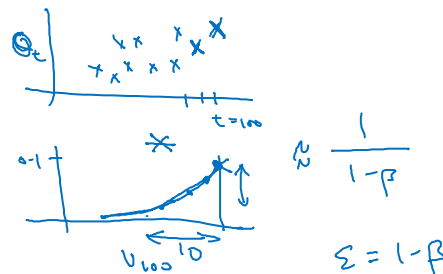
...

$$\begin{aligned} \rightarrow v_{100} &= 0.1\theta_{100} + 0.9 \cancel{v_{99}} (0.1\theta_{99} + 0.9 \cancel{v_{98}}) \\ &= 0.1\theta_{100} + 0.1 \times 0.9 \cdot \theta_{99} + 0.1(0.9)^2\theta_{98} + 0.1(0.9)^3\theta_{97} + 0.1(0.9)^4\theta_{96} \\ &\quad + \dots \\ 0.9^{10} &\approx 0.35 \approx \frac{1}{e} \end{aligned}$$

$$\frac{(1-\epsilon)^{1/\epsilon}}{\epsilon} \approx \frac{1}{e}$$

0.98?

$$\epsilon = 0.02 \rightarrow 0.98^{50} \approx \frac{1}{e}$$



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Implementing exponentially weighted averages

$$v_0 = 0$$

$$v_1 = \beta v_0 + (1 - \beta) \theta_1$$

$$v_2 = \beta v_1 + (1 - \beta) \theta_2$$

$$v_3 = \beta v_2 + (1 - \beta) \theta_3$$

...

$$V_0 := 0$$

$$V_0 := \beta v + (1 - \beta) \theta_1$$

$$V_0 := \beta v + (1 - \beta) \theta_2$$

⋮

$$\rightarrow V_0 = 0$$

Repeat {

Get next θ_t

$$V_0 := \beta V_0 + (1 - \beta) \theta_t \leftarrow$$

}

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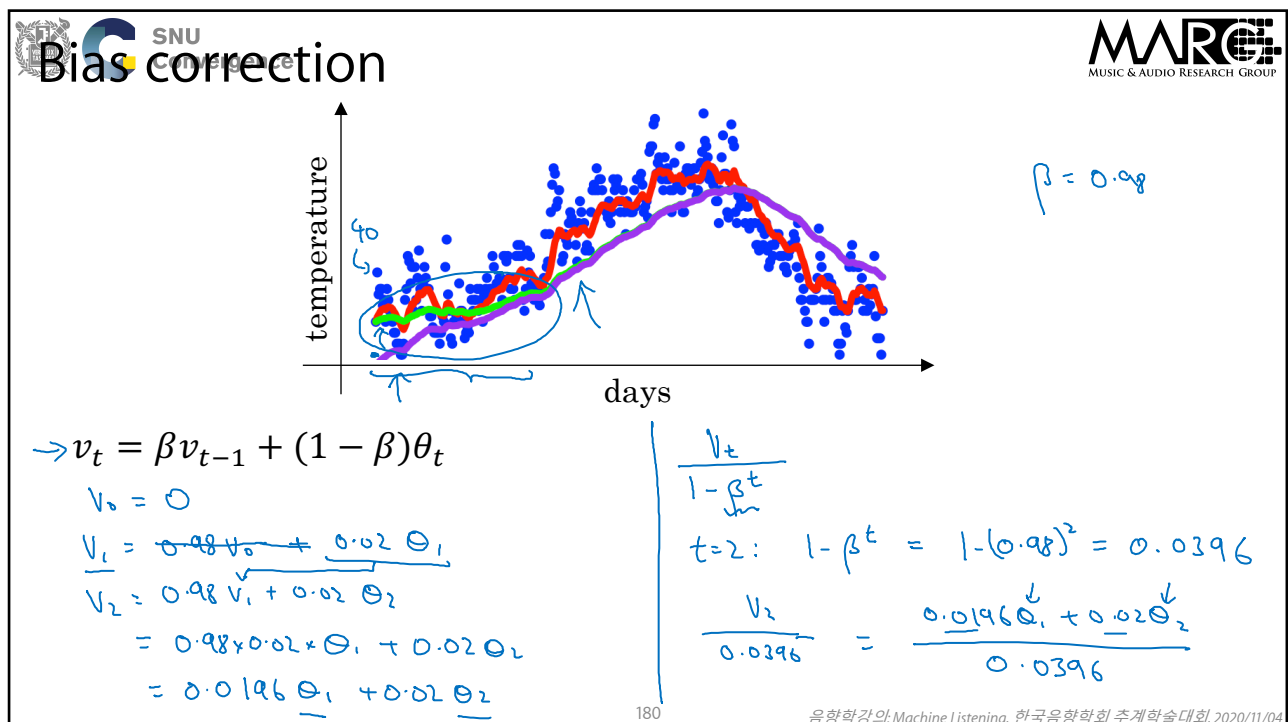


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Optimization Algorithms

Bias correction in exponentially weighted average

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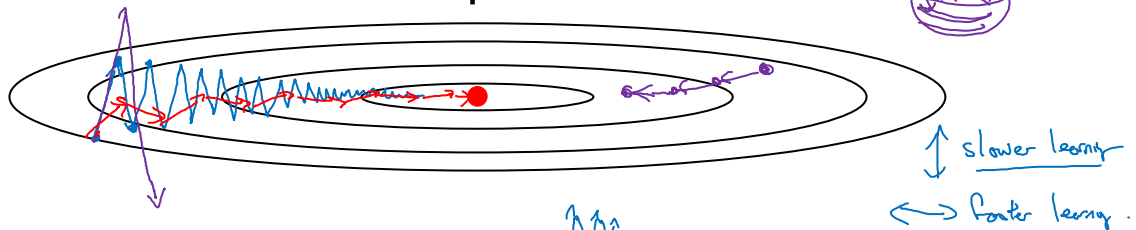
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Optimization Algorithms

Gradient descent with momentum

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Gradient descent example



Momentum:

On iteration t :

Compute $\Delta W, \Delta b$ on current mini-batch.

$$V_{\Delta W} = \beta V_{\Delta W} + (1-\beta) \Delta W$$

$$V_{\Delta b} = \beta V_{\Delta b} + (1-\beta) \Delta b$$

Friction $\rightarrow$ velocity

acceleration

$$W := W - \alpha V_{\Delta W}, \quad b := b - \alpha V_{\Delta b}$$

$$V_0 = \beta V_0 + (1-\beta) \Delta e$$

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Implementation details

$$v_{dw} = 0, v_{db} = 0$$

On iteration t :

Compute dW, db on the current mini-batch

$$\begin{aligned} \rightarrow v_{dw} &= \beta v_{dw} + (1 - \beta) dW \\ \rightarrow v_{db} &= \beta v_{db} + (1 - \beta) db \end{aligned} \quad \left| \quad \begin{aligned} v_{dw} &= \beta v_{dw} + dW \end{aligned} \right. \leftarrow$$

$$W = W - \alpha v_{dw}, \quad b = b - \alpha v_{db}$$

$$\frac{v_{dw}}{1 - \beta^t}$$

Hyperparameters: α, β

$$\beta = 0.9$$

average over last ≈ 10 gradients

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Optimization Algorithms

RMSprop

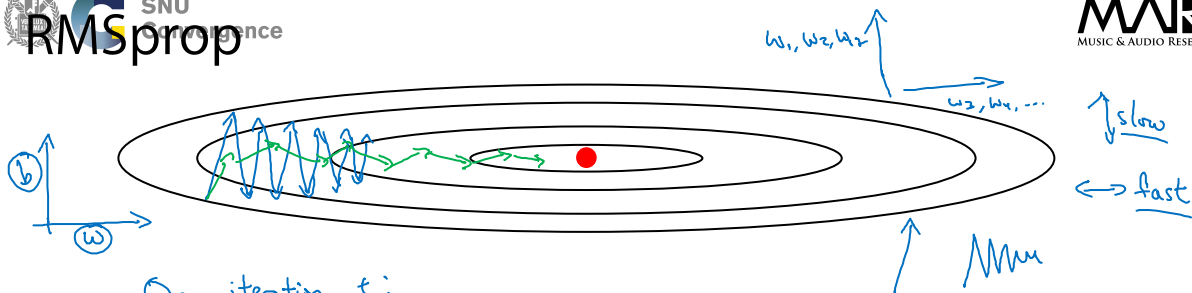
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RMSprop



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On iteration t :

Compute dw, db on current mini-batch

$$\underline{S_{dw}} = \beta_2 \underline{S_{dw}} + (1-\beta_2) \underline{dw^2} \leftarrow \text{small}$$

← element-wise

$$\rightarrow \underline{S_{db}} = \beta_2 \underline{S_{db}} + (1-\beta_2) \underline{db^2} \leftarrow \text{large}$$

$$w := w - \alpha \frac{dw}{\sqrt{\underline{S_{dw}} + \epsilon}} \quad b := b - \alpha \frac{db}{\sqrt{\underline{S_{db}} + \epsilon}}$$

←

$\epsilon = 10^{-8}$

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Optimization Algorithms

Adam optimization algorithm

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Adam optimization algorithm

$$V_{dw} = 0, S_{dw} = 0, V_{db} = 0, S_{db} = 0$$

On iteration t :

Compute dw, db using current mini-batch

$$V_{dw} = \beta_1 V_{dw} + (1 - \beta_1) dw, V_{db} = \beta_1 V_{db} + (1 - \beta_1) db \leftarrow \text{"moment"} \beta_1$$

$$S_{dw} = \beta_2 S_{dw} + (1 - \beta_2) dw^2, S_{db} = \beta_2 S_{db} + (1 - \beta_2) db^2 \leftarrow \text{"RMSprop"} \beta_2$$

$$V_{dw}^{\text{corrected}} = V_{dw} / (1 - \beta_1^t), V_{db}^{\text{corrected}} = V_{db} / (1 - \beta_1^t)$$

$$S_{dw}^{\text{corrected}} = S_{dw} / (1 - \beta_2^t), S_{db}^{\text{corrected}} = S_{db} / (1 - \beta_2^t)$$

$$W := W - \alpha \frac{V_{dw}^{\text{corrected}}}{\sqrt{S_{dw}^{\text{corrected}} + \epsilon}}, b := b - \alpha \frac{V_{db}^{\text{corrected}}}{\sqrt{S_{db}^{\text{corrected}} + \epsilon}}$$

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Hyperparameters choice:

→ α : needs to be tune

→ β_1 : 0.9 → (dw)

→ β_2 : 0.999 → (dw^2)

→ ϵ : 10^{-8}

Adam: Adaptive moment estimation

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Applications of Machine Listening

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Machine Listening Applications

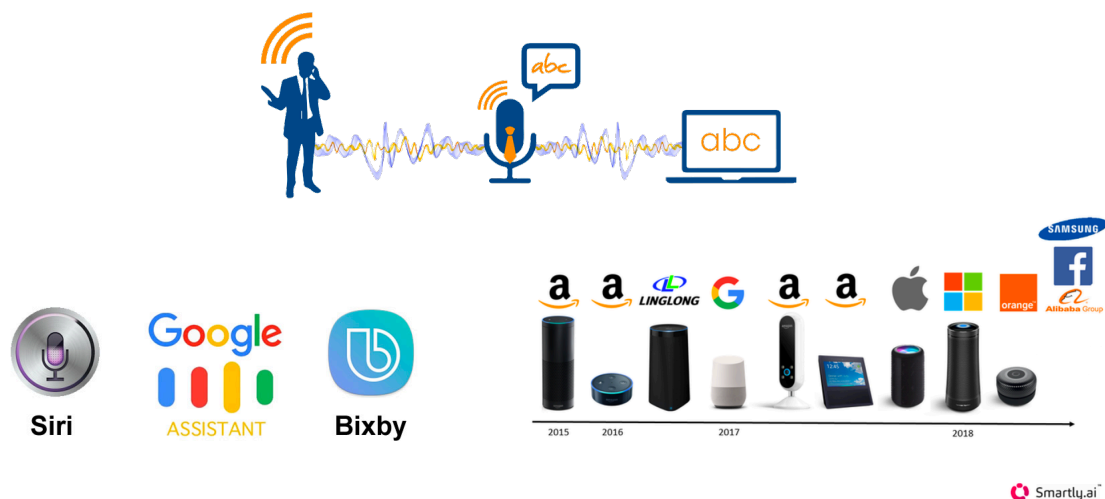
- Speech recognition (speech-to-text or STT)
- Speaker identification/verification
- Emotion recognition
- Speech enhancement
- Source separation
- Automatic music transcription
- Genre/Artist identification
- Music identification
- Lyric-audio alignment
- Speech synthesis (text-to-speech or TTS)
- Singing voice synthesis
- And many more...

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Speech Recognition or STT



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Speech Synthesis or TTS

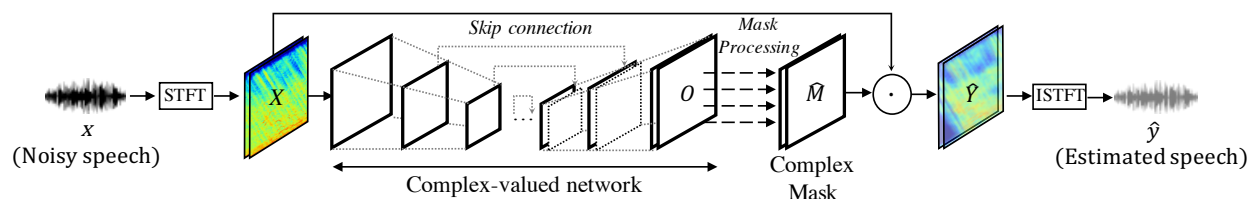


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Speech Enhancement



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Speech Enhancement

영화		
예능		
드라마		
스포츠		
뉴스		

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Speech Enhancement



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Music Identification



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Music Search & Discovery



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Music Source Separation

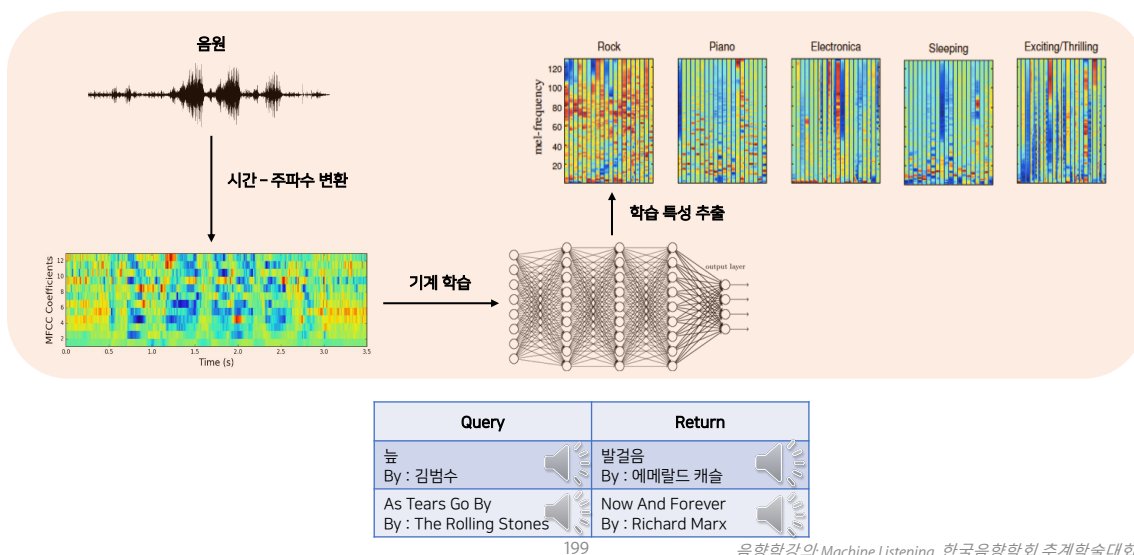
• 오리지널 음악 신호 	• 오리지널 음악 신호 
• 추출된 가창음성 신호 	• 추출된 가창음성 신호 

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Music Recommendation

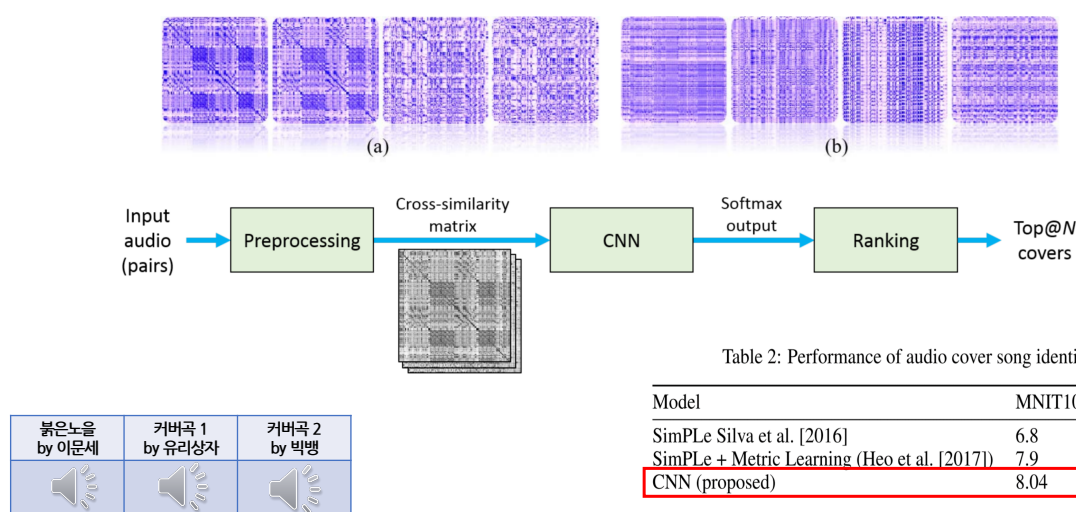


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Audio Coversong Identification



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Voice Cloning



Original actor



Robert Downey Jr.



Deep Fake with voice

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Singing Voice Synthesis

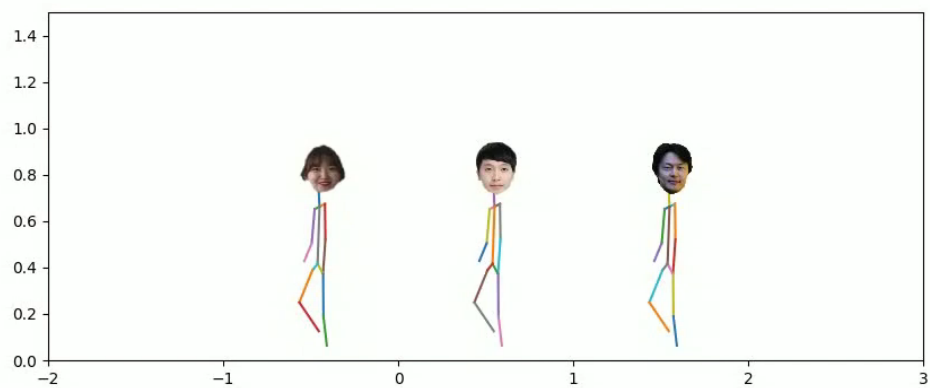


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Listen to Dance

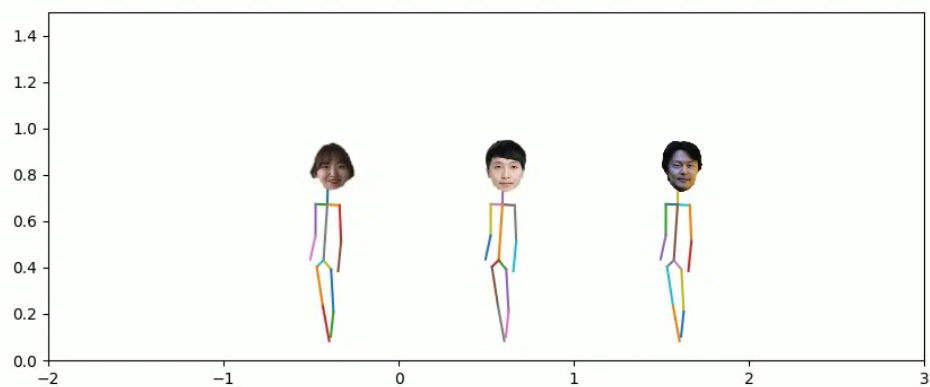


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203

Listen to Dance



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Questions?

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